## SD7226 OVERVIEW

The SD7226 is the industry smallest, a dual-channel, 16-bit, analog-to-digital converter (ADC) supporting sampling rates up to 250MSps. The device uses a multistage pipeline architecture to achieve high signal-to-noise ratio (SNR) and linearity, over wide input signal bandwidth. The SD7226 can be set to operate using either CMOS or LVDS output interface. Programming for configuration and control is accomplished using a 3-wire SPI-compatible serial bus. The digital output data can be programmed to be delivered in offset binary, twos complement format, or gray code.

The device includes a DSP block that features decimation, digital down conversion (DDC), and IQ mismatch correction.

## FEATURES

- 9×9, 3.7mm×3.7mm BGA, with 0.4mm ball-pitch
- SNR: 71.4dBFS at $f_{IN}$ = 70.2MHz and $f_S$ = 250MSps
- SFDR: 75.3dBc at $f_{IN}$ = 70.2MHz and $f_S$ = 250MSps
- -156.4dBFS/Hz input-noise at $f_{IN}$ = 70.2MHz and $f_S$ = 250MSps
- 2.0V nominal input
- Typical power consumption: 631mW at 250MSps
- Integer 1-to-8 input clock divider (1000MHz maximum input rate)
- Sample rates of up to 250MSps
- 1.8V analog supply voltage
- LVDS (ANSI-644 levels) outputs
- Internal ADC voltage reference
- ADC clock duty cycle correction
- Serial port control
- Decimation by 2, 4
- Digital Down-Conversion
- IQ Mismatch Correction
- Energy saving power-down modes

## APPLICATIONS

- Communications
- General-purpose software radios
- I/Q demodulation systems
- Diversity radio systems
- Smart antenna systems
- Multimode digital receivers
- Ultrasound equipment
- Radar/LiDAR applications
- Test and Measurement
- Broadband data applications

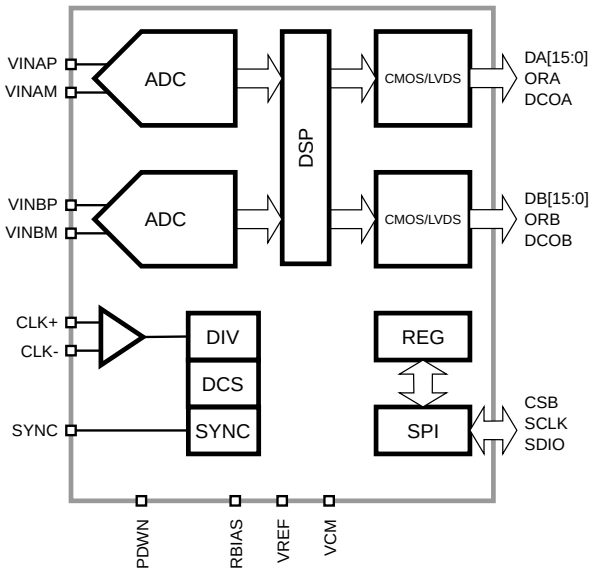



*Figure 1: SD7226 Functional Block Diagram.*

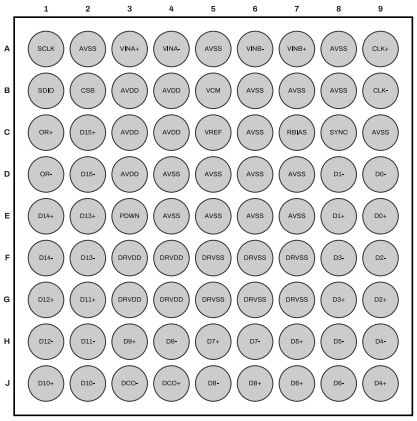


*Figure 2: SD7226 9×9, BGA Package.*

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# SPECIFICATIONS

## DC Specifications

*At $T_A$ = 25°C, $V_{AVDD}$ = 1.8V, $V_{DRVDD}$ = 1.8V, $F_{CLK}$ = 250MHz, $A_{IN}$ = -1dBFS, differential AC-coupled external clock source, LVDS mode, unless otherwise noted.*

*Table 1. DC Specifications.*

| PARAMETER | TEMP | MIN | TYP | MAX | UNIT |
|---|---|---|---|---|---|
| Resolution | | | 16 | | bits |
| Accuracy | | | | | |
| Offset Error | Full | -0.50 | | 0.50 | %FSR |
| Gain Error | Full | | | | %FSR |
| DNL | Full | | | | LSB |
| INL | Full | | | | LSB |
| Matching | | | | | |
| Offset Error | 25°C | -0.80 | | 0.80 | %FSR |
| Gain Error | 25°C | | | | %FSR |
| Temperature Drift | | | | | |
| Offset Error | Full | | +/-0.5 | | ppm/°C |
| Gain Error | Full | | 80.0 | | ppm/°C |
| Internal Voltage Reference | | | | | |
| Output Voltage | Full | 1.02 | | 1.07 | V |
| External Voltage Reference | | | | | |
| Range | Full | 0.90 | | 1.07 | V |
| Input Referred Noise | | | | | |
| $V_{REF}$ = 1.0V | 25°C | | 3.91 | | LSB(rms) |
| Analog Input | | | | | |
| Input Span, $V_{REF}$ = 1.0V | Full | | 2.0 | | V |
| Input Capacitance | Full | | 6.0 | | pF |
| Input Resistance | Full | | 2.0 | | kΩ |
| Input Common-Mode Voltage | Full | | 0.70 | | V |
| Input Common-Mode Range | Full | 0.62 | | 0.93 | V |
| VCM Voltage | Full | | 0.63 | | V |
| VCM Current Capability | Full | | 100 | | µA |
| Reference Input Resistance | Full | | 50 | | kΩ |
| Power Supply | | | | | |
| $V_{AVDD}$ | Full | 1.7 | 1.8 | 1.9 | V |
| $V_{DRVDD}$ | Full | 1.7 | 1.8 | 1.9 | V |
| $I_{AVDD}$ @1.8V | Full | | 280 | | mA |
| $I_{DRVDD}$ @1.8V (LVDS) | Full | | 80 | | mA |
| Power Consumption | | | | | |
| Sine Wave Input (LVDS) | Full | | 631 | | mW |
| Stand-by[1] | Full | | 60 | | mW |
| Power Down | Full | | 10.0 | | mW |

[1] Stand-by power is measured with a sinewave input and active clock.

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## AC Specifications

*At $T_A$ = 25°C, $V_{AVDD}$ = 1.8V, $V_{DRVDD}$ = 1.8V, $F_{CLK}$ = 250MHz, $A_{IN}$ = -1dBFS, differential AC-coupled external clock source, High-Performance Calibration Mode enabled, LVDS mode, unless otherwise noted.*

*Table 2. AC Performance Specifications.*

| PARAMETER | TEMP | MIN | TYP | MAX | UNIT |
|---|---|---|---|---|---|
| Signal-to-Noise Ratio (SNR) | | | | | |
| $f_{IN}$ = 30.2MHz | 25°C | | 71.8 | | dBFS |
| $f_{IN}$ = 70.2MHz | 25°C | | 71.4 | | dBFS |
| | Full | 60.5 | | | dBFS |
| $f_{IN}$ = 140MHz | 25°C | | 70.8 | | dBFS |
| $f_{IN}$ = 220MHz | 25°C | | 68.6 | | dBFS |
| Signal-to-Noise and Distortion Ratio (SNDR) | | | | | |
| $f_{IN}$ = 30.2MHz | 25°C | | 71.1 | | dBFS |
| $f_{IN}$ = 70.2MHz | 25°C | | 70.8 | | dBFS |
| | Full | 60.2 | | | dBFS |
| $f_{IN}$ = 140MHz | 25°C | | 69.6 | | dBFS |
| $f_{IN}$ = 220MHz | 25°C | | 68.2 | | dBFS |
| Effective Number of Bits (ENOB) | | | | | |
| $f_{IN}$ = 30.2MHz | 25°C | | 11.5 | | bits |
| $f_{IN}$ = 70.2MHz | 25°C | | 11.5 | | bits |
| $f_{IN}$ = 140MHz | 25°C | | 11.3 | | bits |
| $f_{IN}$ = 220MHz | 25°C | | 11.0 | | bits |
| Worst 2nd or 3rd Harmonic Power | | | | | |
| $f_{IN}$ = 30.2MHz | 25°C | | -79.1 | | dBc |
| $f_{IN}$ = 70.2MHz | 25°C | | -76.5 | -69.0 | dBc |
| $f_{IN}$ = 140MHz | 25°C | | -74.9 | | dBc |
| $f_{IN}$ = 220MHz | 25°C | | -71.4 | | dBc |
| Worst Non-Harmonic Power | | | | | |
| $f_{IN}$ = 30.2MHz | 25°C | | -78.9 | | dBc |
| $f_{IN}$ = 70.2MHz | 25°C | | -77.3 | -72.0 | dBc |
| $f_{IN}$ = 140MHz | 25°C | | -77.3 | | dBc |
| $f_{IN}$ = 220MHz | 25°C | | -75.1 | | dBc |
| Spurious-Free Dynamic Range[1] (SFDR) | | | | | |
| $f_{IN}$ = 30.2MHz | 25°C | | 76.9 | | dBc |
| $f_{IN}$ = 70.2MHz | 25°C | | 75.3 | | dBc |
| | Full | 69.0 | | | dBc |
| $f_{IN}$ = 140MHz | 25°C | | 74.9 | | dBc |
| $f_{IN}$ = 220MHz | 25°C | | 71.4 | | dBc |
| Two-Tone SFDR | | | | | |
| $f_{IN1}$ = 183.0MHz, $f_{IN2}$ = 187.0MHz | 25°C | | 81.1 | | dBc |
| Crosstalk[2] | 25°C | | -105.0 | | dBc |
| Analog Input Bandwidth | Full | | 650.0 | | MHz |

[1] SNR, SNDR and SFDR exclude the DC and $f_S/2$ bins.

[2] Crosstalk is measured at 100MHz with -1.0dBFS on one channel and no input on the alternate channel.

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## Digital Specification

*$V_{AVDD}$ = 1.8V, $V_{DRVDD}$ = 1.8V, $F_{CLK}$ = 250MHz, $A_{IN}$ = -1dBFS, differential AC-coupled external clock source, DCS disabled, unless otherwise noted.*

*Table 3. Differential Clock Input.*

| PARAMETER | TEMP | MIN | TYP | MAX | UNIT |
|---|---|---|---|---|---|
| Logic Compliance | Full | | CMOS/LVDS/LVPECL | | |
| Internal Common-Mode Bias | Full | | 0.9 | | V |
| Differential Input Voltage | Full | 0.3 | | 1.2 | $V_{p-p}$ |
| Input Voltage Range | Full | $V_{AVSS}$-0.3 | | $V_{AVDD}$+0.2 | V |
| Input Common-Mode Range | Full | 0.75 | | 1.05 | V |
| High Level Input Current | Full | -10 | | 10 | µA |
| Low Level Input Current | Full | -10 | | 10 | µA |
| Input Capacitance | Full | | 1.7 | | pF |
| Input Resistance | Full | | 6.0 | | kΩ |

*Table 4. SYNC Input.*

| PARAMETER | TEMP | MIN | TYP | MAX | UNIT |
|---|---|---|---|---|---|
| Logic Compliance | Full | | CMOS | | |
| Input Voltage Range | Full | $V_{AVSS}$ | | $V_{AVDD}$ | V |
| High Level Input Voltage | Full | 1.22 | | $V_{AVDD}$ | V |
| Low Level Input Voltage | Full | $V_{AGND}$ | | 0.50 | V |
| High Level Input Current | Full | -1 | | 1 | µA |
| Low Level Input Current | Full | -1 | | 1 | µA |
| Input Capacitance | Full | | 1.5 | | pF |
| Input Resistance | Full | | 100 | | kΩ |

*Table 5. CSB Input.*

| PARAMETER | TEMP | MIN | TYP | MAX | UNIT |
|---|---|---|---|---|---|
| High Level Input Voltage | Full | 1.22 | | | V |
| Low Level Input Voltage | Full | | | 0.6 | V |
| High Level Input Current | Full | -1 | | 1 | µA |
| Low Level Input Current | Full | -1 | | 1 | µA |
| Input Capacitance | Full | | 1.5 | | pF |
| Input Resistance | Full | | 100 | | kΩ |

*Table 6. SCLK Input.*

| PARAMETER | TEMP | MIN | TYP | MAX | UNIT |
|---|---|---|---|---|---|
| High Level Input Voltage | Full | 1.22 | | | V |
| Low Level Input Voltage | Full | | | 0.6 | V |
| High Level Input Current | Full | -1 | | 1 | µA |
| Low Level Input Current | Full | -1 | | 1 | µA |
| Input Capacitance | Full | | 1.5 | | pF |
| Input Resistance | Full | | 100 | | kΩ |

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*Table 7. SDIO Input.*

| PARAMETER | TEMP | MIN | TYP | MAX | UNIT |
|---|---|---|---|---|---|
| High Level Input Voltage | Full | 1.22 | | | V |
| Low Level Input Voltage | Full | | | 0.6 | V |
| High Level Input Current | Full | -1 | | 1 | µA |
| Low Level Input Current | Full | -1 | | 1 | µA |
| Input Capacitance | Full | | 1.5 | | pF |
| Input Resistance | Full | | 100 | | kΩ |

*Table 8. PDWN Input.*

| PARAMETER | TEMP | MIN | TYP | MAX | UNIT |
|---|---|---|---|---|---|
| High Level Input Voltage | Full | 1.22 | | | V |
| Low Level Input Voltage | Full | | | 0.6 | V |
| Pull-down Current @1.8V | Full | | | 90 | µA |
| Input Capacitance | Full | | 1.5 | | pF |
| Input Resistance | Full | | 28 | | kΩ |

*Table 9. CMOS Outputs (DATA and OR).*

| PARAMETER | TEMP | MIN | TYP | MAX | UNIT |
|---|---|---|---|---|---|
| High Level Output Voltage @50µA | Full | 1.77 | | | V |
| High Level Output Voltage @0.5mA | Full | 1.75 | | | V |
| Low Level Output Voltage @50µA | Full | | | 0.05 | V |
| Low Level Output Voltage @1.6mA | Full | | | 0.10 | V |

*Table 10. LVDS Outputs (DATA and OR).*

| PARAMETER | TEMP | MIN | TYP | MAX | UNIT |
|---|---|---|---|---|---|
| Differential Output Voltage ($V_{OD}$), ANSI Mode | Full | 250 | 300 | 450 | mV |
| Output Offset Voltage ($V_{OS}$), ANSI Mode | Full | 1.10 | 1.13 | 1.30 | V |
| Differential Output Voltage ($V_{OD}$), Reduced Swing | Full | 150 | 220 | 285 | mV |
| Output Offset Voltage ($V_{OS}$), Reduced Swing | Full | 1.10 | 1.22 | 1.30 | V |

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## Switching Specifications

*$V_{AVDD}$ = 1.8V, $V_{DRVDD}$ = 1.8V, $F_{CLK}$ = 250MHz, $A_{IN}$ = -1dBFS, differential AC-coupled sine wave external clock source, DCS enabled, unless otherwise noted.*

*Table 11. Clock Input Timing.*

| PARAMETER | TEMP | MIN | TYP | MAX | UNIT |
|---|---|---|---|---|---|
| Input Clock Rate | Full | | | 1000 | MHz |
| Conversion Rate (after clock divider) | Full | 20.0 | | 250.0 | MHz |
| *CLK Pulse Width High* ($t_{CH}$) | | | | | |
| Divide-by-1 Mode, DCS Enabled | Full | 1.5 | | | ns |
| Divide-by-1 Mode, DCS Disabled | Full | 1.9 | 2.0 | 2.1 | ns |
| Divide-by-2 Mode Through Divide-by-8 Mode | Full | 0.6 | | | ns |
| Aperture Delay ($t_A$) | Full | | 0.5 | | ns |
| Aperture Uncertainty (Jitter, $t_J$) | Full | | 140.0 | | fs |

*Table 12. SYNC Timing Requirements.*

| PARAMETER | MIN | TYP | MAX | UNIT |
|---|---|---|---|---|
| Set-Up Time ($t_{S,SYNC}$) | | 0.35 | | ns |
| Hold Time ($t_{H,SYNC}$) | | 0.35 | | ns |

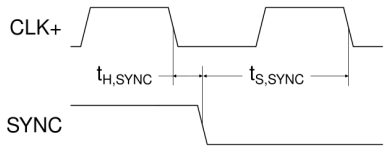



*Figure 2: SYNC Input Timing.*

*Table 13. Data Output.*

| PARAMETER | TEMP | MIN | TYP | MAX | UNIT |
|---|---|---|---|---|---|
| *CMOS Mode*[1] | | | | | |
| Data Propagation Delay ($t_{PD}$) | Full | | 6.0 | | ns |
| DCO Propagation Delay ($t_{DCO}$) | Full | | 7.0 | | ns |
| DCO to Data Skew ($t_{SKEW}$) | Full | -0.9 | -0.5 | 0.0 | ns |
| Pipeline Delay (Latency, L) | Full | | 35.0 | | Cycles |
| *LVDS Mode* | | | | | |
| Data Propagation Delay ($t_{PD}$) | Full | | 8.0 | | ns |
| DCO Propagation Delay ($t_{DCO}$) | Full | | 7.7 | | ns |
| DCO to Data Skew ($t_{SKEW}$) | Full | -0.9 | -0.5 | 0.0 | ns |
| Pipeline Delay (Latency, L) Channel A/Channel B | Full | | 32/32.5 | | Cycles |
| Wake-Up Time (from sleep) | Full | | 5.0 | | µs |
| Wake-Up Time (from power down) | Full | | 250.0 | | µs |
| Out-of-Range Recovery Time | Full | | 3 | | Cycles |

[1] Up to 125MHz.

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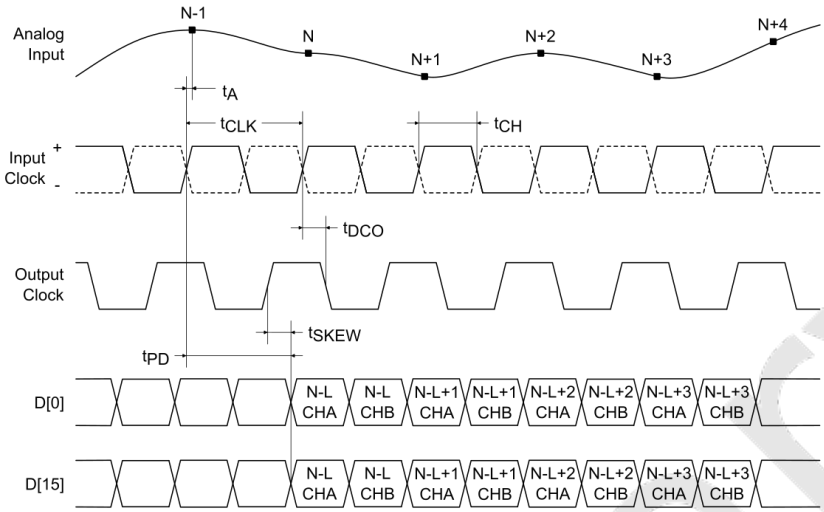



*Figure 3: Parallel LVDS Output Mode Timing.*

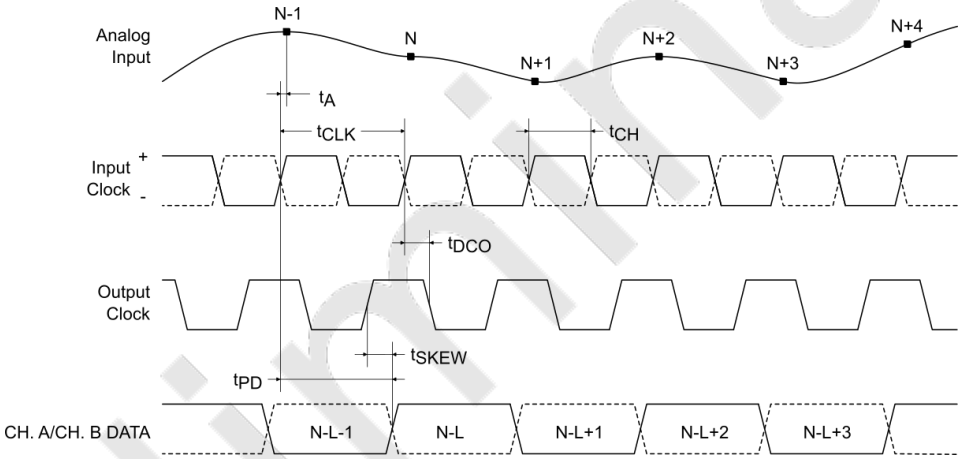



*Figure 4: Parallel CMOS Output Mode Timing.*

## Output Modes

The SD7226 supports both CMOS and LVDS interfaces. Two LVDS samples are sent out every clock cycle using both the rising and the falling edge of the output clock. In interleaved parallel mode, the two ADCs share the full set of output pins. The first half-cycle contains ADC A data and the second ADC B data.

## LVDS→CMOS Conversion

The output interface can be converted to CMOS using the register writes in Table 14, where each IO pad is programmed to a CMOS driver.

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*Table 14. Output Mode Conversion (LVDS→CMOS).*

| Register | Value | Register | Value | Register | Value | Register | Value |
|---|---|---|---|---|---|---|---|
| 0x401 | *0xc3* | 0x403 | *0xe1* | 0x405 | *0x4e0* | 0x407 | *0x4e0* |
| 0x409 | *0x8e0* | 0x40b | *0x8e0* | 0x40d | *0x4e0* | 0x40f | *0x4e0* |
| 0x411 | *0x4e0* | 0x413 | *0x4e0* | 0x415 | *0x8e0* | 0x417 | *0x8e0* |
| 0x419 | *0x4e0* | 0x41b | *0x4e0* | 0x41d | *0x8e0* | 0x41f | *0x8e0* |
| 0x421 | *0x8e0* | 0x423 | *0x8e0* | 0x425 | *0x4e0* | 0x427 | *0x10e0* |
| 0x429 | *0x8e0* | 0x42b | *0x4e0* | 0x42d | *0x8e0* | 0x42f | *0x8e0* |
| 0x431 | *0x8e0* | 0x433 | *0x8e0* | 0x435 | *0xce0* | 0x437 | *0xce0* |
| 0x439 | *0x4e0* | 0x43b | *0xce0* | 0x43d | *0xce0* | 0x43f | *0xce0* |
| 0x441 | *0x4e0* | 0x443 | *0x4e0* | 0x445 | *0x4e0* | 0x447 | *0x4e0* |
| 0x449 | *0x8e0* | 0x44b | *0x8e0* | 0x44d | *0xc3* | 0x44f | *0xc3* |
| 0x451 | *0xc3* | 0x453 | *0xc3* | 0x466 bit[0] | *0x0* | 0x466 bit[2] | *0x0* |

The LVDS interface is recommended when the output clock rate exceeds 125MHz.

## Output Timing Control

Data delay control, bits [8:5], in register *0x473*, can be used to move the CMOS data relative to output clock. Bit [9] inverts the internal clock sending the data out. Bits [13:10], in register *0x473*, control the output clock delay relative to output data, while bit [14] inverts the clock. The delay step size is about 250ps.

Data delay control is not available in interleaved or multiplexed output mode.

Driving long traces or large load capacitance with CMOS output produces voltage ripple in the digital IO supply and ground, which may couple on chip to the sensitive analog circuits in the ADC and degrade the performance. This effect can be reduced by using an external buffer IC. The use of LVDS output interface largely eliminates issues with IO supply noise.

Due to the close proximity of the ADC_B LSB pin to the clock input pins, there is potential for noise coupling, which can degrade jitter performance. To mitigate this, the PCB layout should ensure that traces to these pins are routed separately, with a grounded shield placed between them. Jitter impact depends strongly on how data transitions aling with sampling clock edges. Data delay control can be used to move the CMOS data transitions away from sensitive sampling clock edge.

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## ABSOLUTE MAXIMUM RATINGS

*Table 15. Absolute Maximum Ratings ($T_A$ = 25°C, unless otherwise specified).*

| Parameter | Symbol | Conditions | Min | Max | Units |
|---|---|---|---|---|---|
| AVDD | $V_{AVDD}$ | Relative to AVSS | -0.3 | 2.0 | V |
| DRVDD | $V_{DRVDD}$ | Relative to AVSS | -0.3 | 2.0 | V |
| VINAP/VINBP, VINAM/VINBM | | Relative to AVSS | -0.3 | 2.0 | V |
| CLK+, CLK- | | Relative to AVSS | -0.3 | 2.0 | V |
| SYNC | | Relative to AVSS | -0.3 | 2.0 | V |
| RBIAS | | Relative to AVSS | -0.3 | 2.0 | V |
| VCM | | Relative to AVSS | -0.3 | 2.0 | V |
| VREF | | Relative to AVSS | -0.3 | 2.0 | V |
| CSB | | Relative to AVSS | -0.3 | 2.0 | V |
| SCLK | | Relative to AVSS | -0.3 | 2.0 | V |
| SDIO | | Relative to AVSS | -0.3 | 2.0 | V |
| PDWN | | Relative to AVSS | -0.3 | 2.0 | V |
| DA0, ..., DA15<br>DB0, ..., DB15 | | Relative to AVSS | -0.3 | 2.0 | V |
| DCO+, DCO- | | Relative to AVSS | -0.3 | 2.0 | V |
| Operating Temperature Range (Ambient) | | | -40 | 85 | °C |
| Maximum Junction Temperature Under Bias | | | | 125 | |
| Storage Temperature Range (Ambient) | | | -65 | 150 | |

Notes:

1. Stresses beyond those listed under Table 15 may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

**ESD CAUTION**.

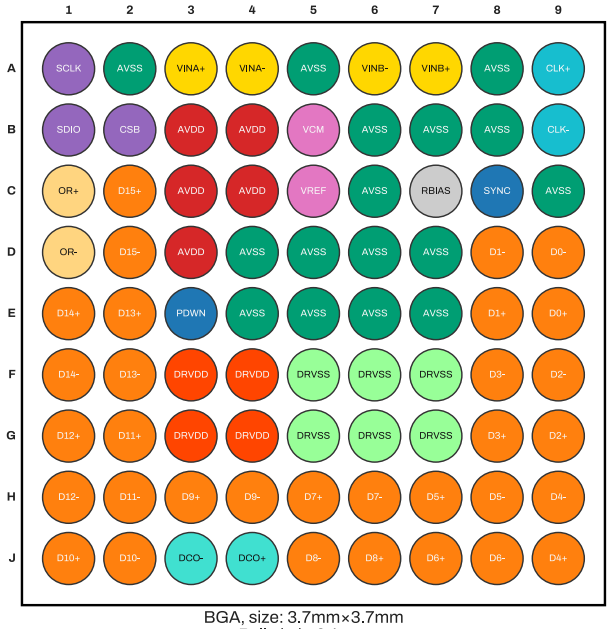


**Electrostatic Discharge Sensitive Device.**

Proper ESD precautions should be observed to prevent performance degradation or loss of functionality.

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## PACKAGE




BGA, size: 3.7mm×3.7mm
Ball pitch: 0.4mm

*Figure 5: SD7226 Package Top View for Parallel LVDS Configuration.*

*Table 16. Pin Descriptions for Parallel LVDS Configuration.*

| Number | Name | Type | Comment |
|---|---|---|---|
| A1 | SCLK | Input | Serial interface clock input. |
| A2, A5, A8, B6, B7, B8, C6, C9, D4, D5, D6, D7, E4, E5, E6, E7 | AVSS | Ground | Analog Ground. The exposed thermal paddle on the bottom of the package provides the analog ground for the part. The exposed paddle must be connected to ground for proper operation. |
| A3 | VINA+ | Input | Differential Analog Input Pin (Plus) for Channel A. |
| A4 | VINA- | Input | Differential Analog Input Pin (Minus) for Channel A. |
| A6 | VINB- | Input | Differential Analog Input Pin (Minus) for Channel B. |
| A7 | VINB+ | Input | Differential Analog Input Pin (Plus) for Channel B. |
| A9 | CLK+ | Input | ADC Clock Input (Plus). |
| B1 | SDIO | InOut | SPI Serial Data I/O. |
| B2 | CSB | Input | SPI Chip Select (Active Low). |
| B3, B4, C3, C4, D3 | AVDD | Power | Analog Power Supply (1.8 V Nominal). |
| B5 | VCM | Output | This pin outputs the common-mode voltage that can be used externally to bias the analog input pins. |
| B9 | CLK- | Input | ADC Clock Input (Minus). |
| C1 | OR+ | Output | Channel A/Channel B LVDS Overrange (Plus). |
| C2 | D15+ | Output | Channel A/Channel B LVDS Output Data 15 (Plus). |
| C5 | VREF | InOut | Voltage Reference Input/Output. |

| Number | Name | Type | Comment |
|---|---|---|---|
| C7 | RBIAS | InOut | External Reference Bias Resistor. Connect to 30 kΩ (1% tolerance) resistor to ground. |
| C8 | SYNC | Input | Digital Synchronization Pin. |
| D1 | OR- | Output | Channel A/Channel B LVDS Overrange (Minus). |
| D2 | D15- | Output | Channel A/Channel B LVDS Output Data 15 (Minus). |
| D8 | D1- | Output | Channel A/Channel B LVDS Output Data 1 (Minus). |
| D9 | D0- | Output | Channel A/Channel B LVDS Output Data 0 (Minus). |
| E1 | D14+ | Output | Channel A/Channel B LVDS Output Data 14 (Plus). |
| E2 | D13+ | Output | Channel A/Channel B LVDS Output Data 13 (Plus). |
| E3 | PDWN | Input | Power-Down Input (Active High). The operation of this pin depends on the SPI mode and can be configured as power-down or stand-by. |
| E8 | D1+ | Output | Channel A/Channel B LVDS Output Data 1 (Plus). |
| E9 | D0+ | Output | Channel A/Channel B LVDS Output Data 0 (Plus). |
| F1 | D14- | Output | Channel A/Channel B LVDS Output Data 14 (Minus). |
| F2 | D13- | Output | Channel A/Channel B LVDS Output Data 13 (Minus). |
| F3, F4, G3, G4 | DRVDD | Power | Digital I/O Supply. |
| F5, F6, F7, G5, G6, G7 | DRVSS | Ground | Digital I/O Ground. |
| F8 | D3- | Output | Channel A/Channel B LVDS Output Data 3 (Minus). |
| F9 | D2- | Output | Channel A/Channel B LVDS Output Data 2 (Minus). |
| G1 | D12+ | Output | Channel A/Channel B LVDS Output Data 12 (Plus). |
| G2 | D11+ | Output | Channel A/Channel B LVDS Output Data 11 (Plus). |
| G8 | D3+ | Output | Channel A/Channel B LVDS Output Data 3 (Plus). |
| G9 | D2+ | Output | Channel A/Channel B LVDS Output Data 2 (Plus). |
| H1 | D12- | Output | Channel A/Channel B LVDS Output Data 12 (Minus). |
| H2 | D11- | Output | Channel A/Channel B LVDS Output Data 11 (Minus). |
| H3 | D9+ | Output | Channel A/Channel B LVDS Output Data 9 (Plus). |
| H4 | D9- | Output | Channel A/Channel B LVDS Output Data 9 (Minus). |
| H5 | D7+ | Output | Channel A/Channel B LVDS Output Data 7 (Plus). |
| H6 | D7- | Output | Channel A/Channel B LVDS Output Data 7 (Minus). |
| H7 | D5+ | Output | Channel A/Channel B LVDS Output Data 5 (Plus). |
| H8 | D5- | Output | Channel A/Channel B LVDS Output Data 5 (Minus). |
| H9 | D4- | Output | Channel A/Channel B LVDS Output Data 4 (Minus). |
| J1 | D10+ | Output | Channel A/Channel B LVDS Output Data 10 (Plus). |
| J2 | D10- | Output | Channel A/Channel B LVDS Output Data 10 (Minus). |
| J3 | DCO- | Output | LVDS Data Clock Output (Minus). |
| J4 | DCO+ | Output | LVDS Data Clock Output (Plus). |
| J5 | D8- | Output | Channel A/Channel B LVDS Output Data 8 (Minus). |
| J6 | D8+ | Output | Channel A/Channel B LVDS Output Data 8 (Plus). |
| J7 | D6+ | Output | Channel A/Channel B LVDS Output Data 6 (Plus). |
| J8 | D6- | Output | Channel A/Channel B LVDS Output Data 6 (Minus). |
| J9 | D4+ | Output | Channel A/Channel B LVDS Output Data 4 (Plus). |

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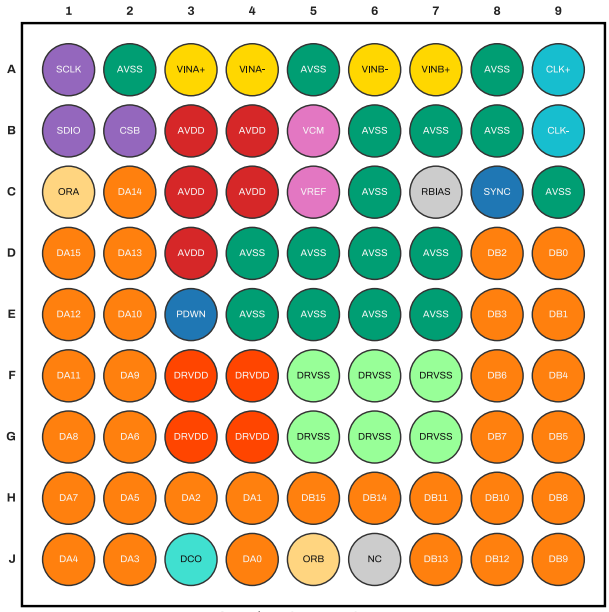



*Figure 6: SD7226 Package Top View for Parallel CMOS Configuration.*

*Table 17. Pin Descriptions for Parallel CMOS Configuration.*

| Number | Name | Type | Comment |
|---|---|---|---|
| A1 | SCLK | Input | Serial interface clock input. |
| A2, A5, A8, B6, B7, B8, C6, C9, D4, D5, D6, D7, E4, E5, E6, E7 | AVSS | Ground | Analog Ground. The exposed thermal paddle on the bottom of the package provides the analog ground for the part. The exposed paddle must be connected to ground for proper operation. |
| A3 | VINA+ | Input | Differential Analog Input Pin (Plus) for Channel A. |
| A4 | VINA- | Input | Differential Analog Input Pin (Minus) for Channel A. |
| A6 | VINB- | Input | Differential Analog Input Pin (Minus) for Channel B. |
| A7 | VINB+ | Input | Differential Analog Input Pin (Plus) for Channel B. |
| A9 | CLK+ | Input | ADC Clock Input (Plus). |
| B1 | SDIO | InOut | SPI Serial Data I/O. |
| B2 | CSB | Input | SPI Chip Select (Active Low). |
| B3, B4, C3, C4, D3 | AVDD | Power | Analog Power Supply (1.8 V Nominal). |
| B5 | VCM | Output | This pin outputs the common-mode voltage that can be used externally to bias the analog input pins. |
| B9 | CLK- | Input | ADC Clock Input (Minus). |
| C1 | ORA | Output | Channel A Overrange Output. |
| C2 | DA14 | Output | Channel A CMOS Output Data 14. |
| C5 | VREF | InOut | Voltage Reference Input/Output. |

| Number | Name | Type | Comment |
|---|---|---|---|
| C7 | RBIAS | InOut | External Reference Bias Resistor. Connect to 30 kΩ (1% tolerance) resistor to ground. |
| C8 | SYNC | Input | Digital Synchronization Pin. |
| D1 | DA15 | Output | Channel A CMOS Output Data 15. |
| D2 | DA13 | Output | Channel A CMOS Output Data 13. |
| D8 | DB2 | Output | Channel B CMOS Output Data 2. |
| D9 | DB0 | Output | Channel B CMOS Output Data 0 (LSB). |
| E1 | DA12 | Output | Channel A CMOS Output Data 12. |
| E2 | DA10 | Output | Channel A CMOS Output Data 10. |
| E3 | PDWN | Input | Power-Down Input (Active High). The operation of this pin depends on the SPI mode and can be configured as power-down or stand-by. |
| E8 | DB3 | Output | Channel B CMOS Output Data 3. |
| E9 | DB1 | Output | Channel B CMOS Output Data 1. |
| F1 | DA11 | Output | Channel A CMOS Output Data 11. |
| F2 | DA9 | Output | Channel A CMOS Output Data 9. |
| F3, F4, G3, G4 | DRVDD | Power | Digital I/O Supply. |
| F5, F6, F7, G5, G6, G7 | DRVSS | Ground | Digital I/O Ground. |
| F8 | DB6 | Output | Channel B CMOS Output Data 6. |
| F9 | DB4 | Output | Channel B CMOS Output Data 4. |
| G1 | DA8 | Output | Channel A CMOS Output Data 8. |
| G2 | DA6 | Output | Channel A CMOS Output Data 6. |
| G8 | DB7 | Output | Channel B CMOS Output Data 7. |
| G9 | DB5 | Output | Channel B CMOS Output Data 5. |
| H1 | DA7 | Output | Channel A CMOS Output Data 7. |
| H2 | DA5 | Output | Channel A CMOS Output Data 5. |
| H3 | DA2 | Output | Channel A CMOS Output Data 2. |
| H4 | DA1 | Output | Channel A CMOS Output Data 1. |
| H5 | DB15 | Output | Channel B CMOS Output Data 15. |
| H6 | DB14 | Output | Channel B CMOS Output Data 14. |
| H7 | DB11 | Output | Channel B CMOS Output Data 11. |
| H8 | DB10 | Output | Channel B CMOS Output Data 10. |
| H9 | DB8 | Output | Channel B CMOS Output Data 8. |
| J1 | DA4 | Output | Channel A CMOS Output Data 4. |
| J2 | DA3 | Output | Channel A CMOS Output Data 3. |
| J3 | DCO | Output | Clock Output. |
| J4 | DA0 | Output | Channel A CMOS Output Data 0 (LSB). |
| J5 | ORB | Output | Channel B Overrange Output. |
| J6 | NC | | Do not connect. |
| J7 | DB13 | Output | Channel B CMOS Output Data 13. |
| J8 | DB12 | Output | Channel B CMOS Output Data 12. |
| J9 | DB9 | Output | Channel B CMOS Output Data 9. |

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## TYPICAL PERFORMANCE CHARACTERISTICS

*At $T_A$ = 25°C, $V_{AVDD}$ = 1.8V, $V_{DRVDD}$ = 1.8V, $F_{CLK}$ = 250MHz, $A_{IN}$ = -1dBFS, differential AC-coupled clock source, High-Performance Calibration Mode enabled, LVDS mode, unless otherwise noted.*

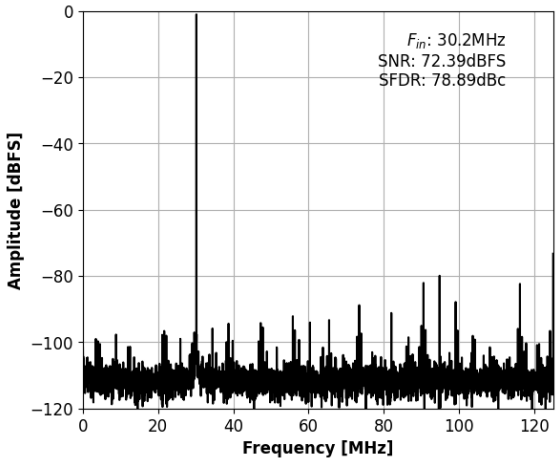



*Figure 7: Single-Tone FFT with $f_{IN}$=30.2MHz.*

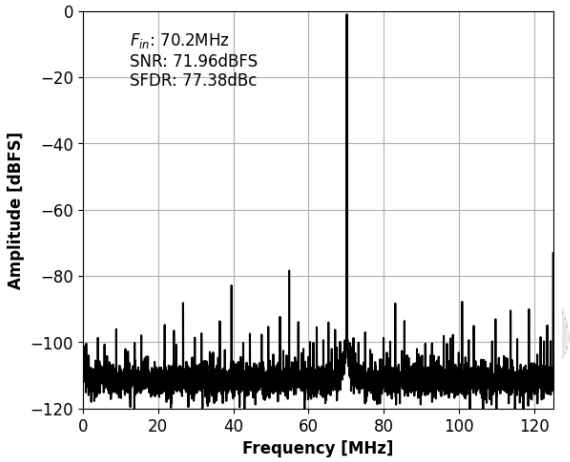



*Figure 8: Single-Tone FFT with $f_{IN}$=70.2MHz.*

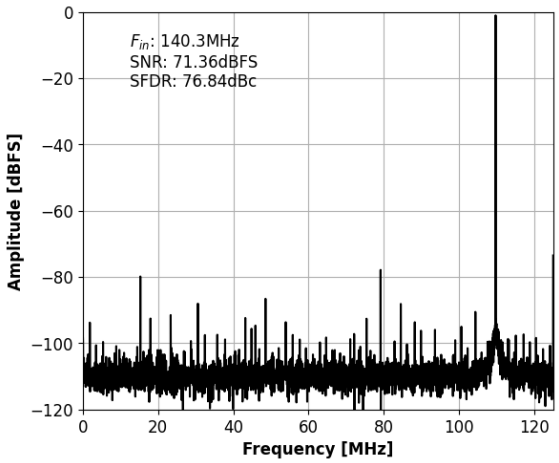



*Figure 9: Single-Tone FFT with $f_{IN}$=140MHz.*

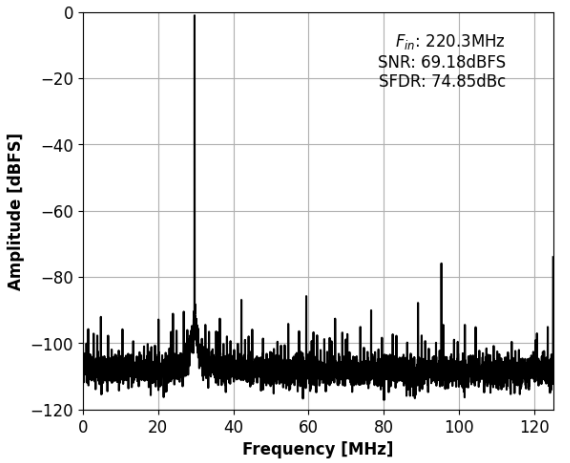



*Figure 10: Single-Tone FFT with $f_{IN}$=220MHz.*

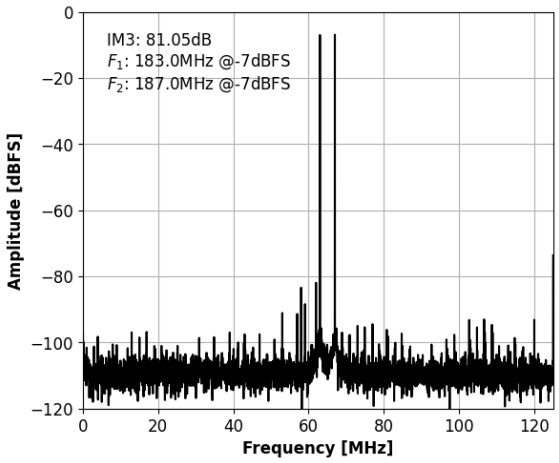



*Figure 11: Two-Tone FFT with $f_{IN1}$=183.0MHz, $f_{IN2}$=187.0MHz.*

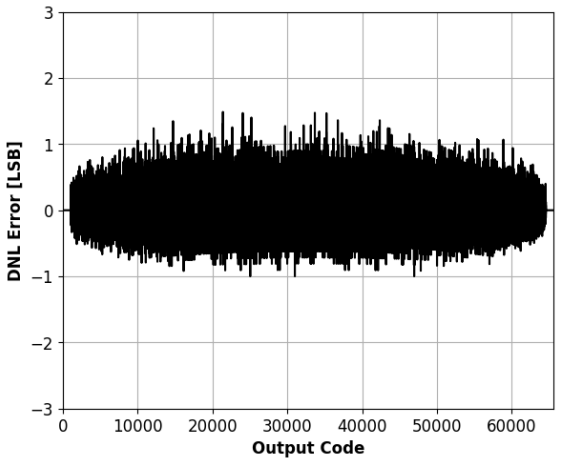



*Figure 12: DNL Error with $f_{IN}$=6.55MHz.*

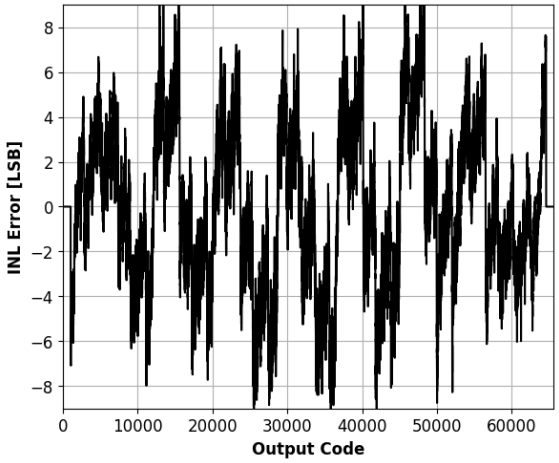



Figure 13: INL Error with $f_{IN}$=6.55MHz.

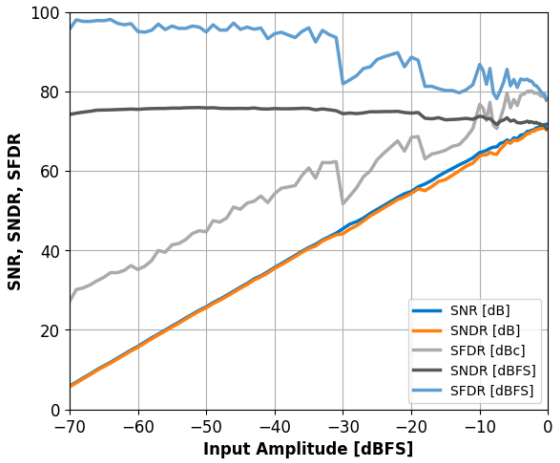



Figure 14: Single-Tone SNR, SNDR and SFDR vs. Input Amplitude with $f_{IN}$=70.25MHz.

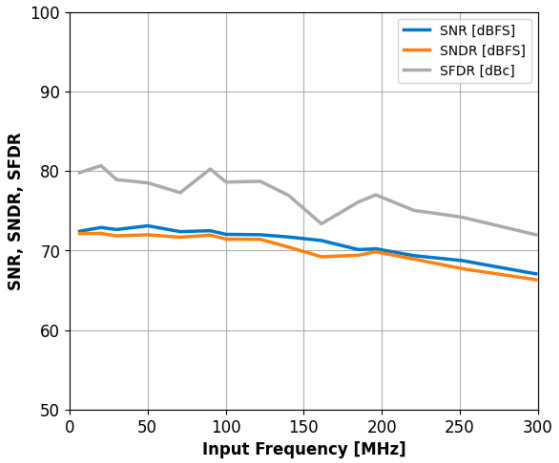



Figure 15: Single-Tone SNR, SNDR and SFDR vs. Input Frequency with $f_S$=250MHz.

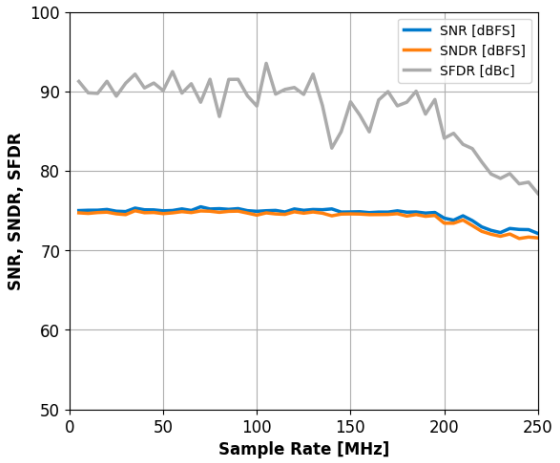



Figure 16: Single-Tone SNR, SNDR and SFDR vs. Sample Rate with $f_{IN}$=70.25MHz.

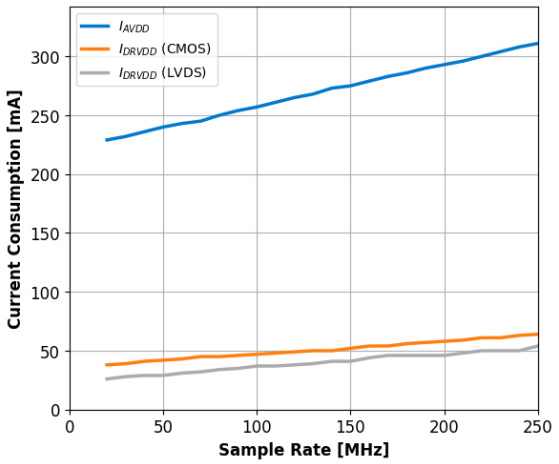



Figure 17: Current vs. Sample Rate.

## EQUIVALENT CIRCUITS

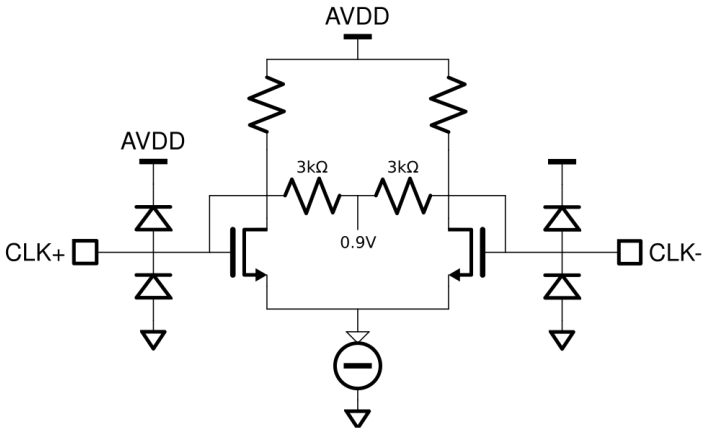



*Figure 18: Equivalent Clock Input Circuit.*

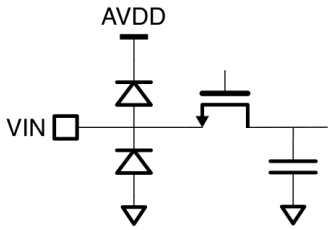



*Figure 19: Equivalent Analog Input Circuit.*

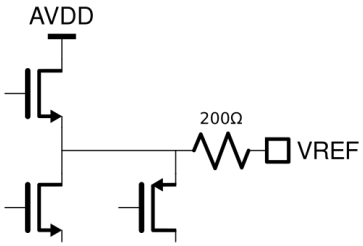



*Figure 20: Equivalent VREF Circuit.*

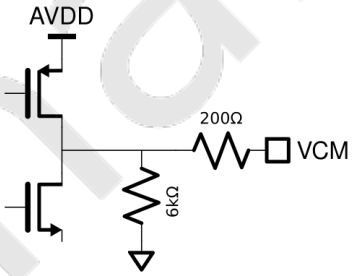



*Figure 21: Equivalent VCM Circuit.*

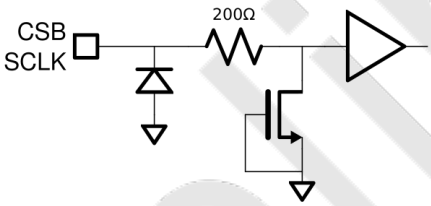



*Figure 22: Equivalent CSB or SCLK Input Circuit.*

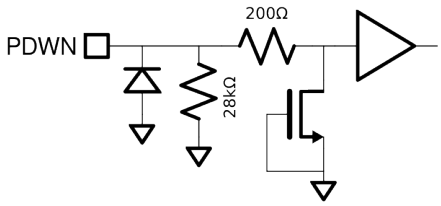



*Figure 23: Equivalent PDWN Input Circuit.*

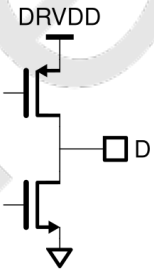



*Figure 24: Equivalent Digital Output Circuit.*

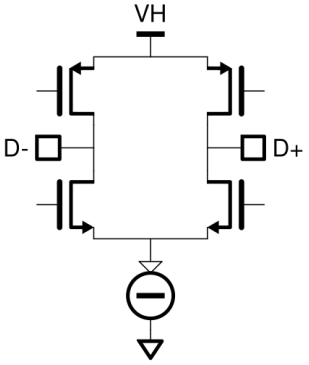



*Figure 25: Equivalent LVDS output Circuit.*

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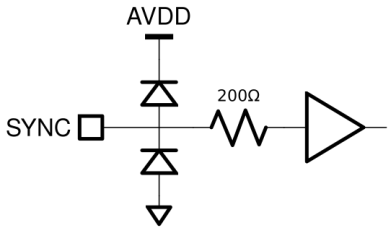



*Figure 26: Equivalent SYNC Input Circuit.*

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# THEORY OF OPERATION

## ADC Architecture

The ADC uses a pipelined architecture and innovative patented switched-capacitor circuits. Its fully differential design provides exceptional immunity to power supply noise and minimizes reference voltage self-modulation. A built-in Sample-and-Hold (S/H) function is integrated into the input stage of the pipeline structure.

## Analog Input

The input stage of the ADC behaves as a switched-capacitor network, presenting itself to the driving circuit as a combination of a switch and a sampling capacitor. The capacitor is reset prior to each conversion cycle, effectively eliminating non-linear memory effects commonly observed in some pipelined ADC architectures. The ADC does not include an internal common-mode bias therefore, the driving source must provide an appropiate common-mode voltage.

## Differential Clock Input

The SD7226 features a differential clock receiver with an integrated common-mode bias. For proper operation, the clock inputs should be AC-coupled using 10nF capacitors.

**Differential Clock Configuration.** For optimal jitter performance, a differential clock source is recommended. The differential clock signals to CLK+ and CLK- should be connected through a 10nF AC-coupling capacitors.

**Single-Ended Clock Configuration.** If a single-ended clock source is used, the signal source should be AC-coupled to the CLK+ pin. In this configuration, a 10nF capacitor should be connected between the CLK- pin and analog ground to maintain proper biasing.

**Clock Jitter Considerations.** Clock jitter has a significant impact on the ADC's signal-to-noise ratio (SNR). The sensitivity to jitter increases with input signal frequency. For best performance, a low-noise differential clock with fast edge transitions should be used.

**Note**: The part can enter in manufacturing test mode if both differential clock inputs are held low for more than 1ms. This behavior can be prevented by writing 1 to register *0x4cf*, bit[15].

## Clock Divider

The ADC includes a programmable clock divider that allows the input clock to be divided by integer values from 2 to 8. The divider is configured by setting register *0x463*, bits [7:5], to the desired division value. A value of 0 (default) bypasses the divider entirely.

By default, the divider starts asynchronously. If the application requires synchronization across multiple ADC devices, the SYNC pin provides means for aligning clock dividers to guarantee sampling on the same clock edge. If clock divider is bypassed, this feature is not needed.

Synchronization is enabled by setting register *0x463*, bit [10], to 1. When the bit is 0, SYNC pin is ignored. SYNC is a level sensitive, active high, reset pin that is captured on the falling edge of the input clock. As long as SYNC is high, the divider stays in reset. It starts dividing on the first falling clock edge after SYNC goes low.

Make sure that the falling edge of SYNC meets the specified setup and hold times relative to the input clock, as shown in Table 12. Pay attention to the length matching of the clock and SYNC routing to the multiple ADC chips. For test and debugging purposes, the level of received and captured SYNC signal can be read from register *0x4ff*, bit [10].

**Clock Duty-Cycle Requirements.** The ADC utilizes both the rising and falling edges of the input clock (or the divided clock, if the divider is enabled) for internal sampling operations. To achieve optimal performance, especially at the maximum sampling rate, the clock duty cycle should be as close to 50% as possible.

When the divider is enabled with an even divisor, a 50% duty cycle is guaranteed by design. If the application uses a clock with a non-ideal duty cycle, a Duty Cycle Stabilizer (DCS) can be enabled to improve performance.

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## ADC Self-Calibration

The ADC includes an automatic calibration mechanism that is executed at power-up to ensure optimal performance. It calibrates out capacitor mismatch and the effects of operational amplifier finite gain and bandwidth. For best results, calibration should be performed at the actual sampling rate used during operation. To support this, the device continuously monitors the sampling clock frequency and automatically re-triggers calibration if a significant change is detected. This feature can be disabled by setting register *0x4c9*, bit[0], to 1.

Manual calibration can also be initiated by toggling both *0xdc1*, bit[4] and *0xcc1*, bit[4] from 0 to 1. The calibration process is implemented using a state-machine architecture, ensuring a deterministic and predictable calibration time. By default the calibration takes 46 million ADC sampling clock cycles. Calibration completion status can be read from *0xdef*, bit[0], for the first ADC and *0xcef*, bit[0], for the second ADC after the status readback is enabled by writing 0x1c to both *0xdeb* and *0xceb*. Bit value 1 indicates that calibration is complete.

In addition, the ADC supports a Background Calibration (BGC) mode, which is disabled by default. When enabled, parameters for compensating changes in operational amplifier gain and bandwidth are continuously updated. BGC is beneficial in environments with large temperature variations near the hot end of the specified operating temperature range. For its operation, BGC uses a dither signal which is injected into the ADC input signal path. This dither consumes approximately 0.8dB of the ADC's input range, resulting in earlier clipping compared to when BGC is disabled. When BGC is enabled calibration completion status bit remains 0.

To enable BGC:

- Write 0b01 to register 0xde1, bits [15:14]
- Write 0b01 to register 0xce1, bits [15:14]
- Write 0x0d80 to register 0xdcf
- Write 0x0d80 to register 0xccf
- Write 0b0 to register 0xde3, bit [3]
- Write 0b0 to register 0xce3, bit [3]

Calibration parameters can be altered to optimize ADC performance and the duration of the calibration according to Table 18.

High Performance settings improve the low frequency SFDR compared to the default settings while the High Speed Settings minimize the calibration time at the cost of small SFDR and SNDR degradation. If the parameters are altered, the calibration has to be reinitiated as described earlier.

*Table 18. Calibration Settings.*

| Register Address | Default Value | High Performance Value | High Speed Value |
|---|---|---|---|
| 0xccd | *0x1483* | *0x3483* | *0x348c* |
| 0xcd5 | *0x1483* | *0x3483* | *0x348c* |
| 0xcdd | *0x16c3* | *0x36c3* | *0x36cc* |
| 0xce1 | *0xbbe8* | *0xbb80* | *0xbb20* |
| 0xdcd | *0x1483* | *0x3483* | *0x348c* |
| 0xdd5 | *0x1483* | *0x3483* | *0x348c* |
| 0xddd | *0x16c3* | *0x36c3* | *0x36cc* |
| 0xde1 | *0xbbe8* | *0xbb80* | *0xbb20* |
| Duration (clock cycles) | 46M | 25M | 6.3M |

## Stand-by and Power-Down Modes

The SD7226 supports two power-saving modes, power down and stand-by, that can be used when the ADCs are not actively sampling. In both modes, the SPI interface (if enabled) remains operational.

**Power-Down Mode.** This mode disables most of the internal circuitry, resulting in the lowest residual supply current. It is

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ideal for applications requiring minimal power consumption during idle periods. Power-down mode can be enabled via:

- The PDWN pin when operating in the external pin mode, or
- Setting register *0x457*, bit[7], to 1.

**Stand-by Mode.** Stand-by mode offers a faster wake-up time compared to power-down mode, at the cost of slightly higher residual current. It is suitable for applications that require rapid recovery from idle states. To enable stand-by mode, set register *0x457*, bits [3] and [8], to 1.

**Note**: It is recommended to disable the ADC auto-calibration when using stand-by mode to avoid unintended calibration cycles during transitions.

## Commonly Used Functions

The SD7226 offers several functions available via dual function pin controls (external pin mode vs. SPI mode). The SD7226 detects the SPI mode during the first SPI transaction after power-up. If the user does not want to program the device via the SPI interface, the dual function is available.

**Duty-Cycle Stabilizer**. The Duty-Cycle Stabilizer (DCS) retimes the falling edge of the clock thus providing an internal clock signal with a 50% duty cycle. The DCS is useful in systems where the input clock has an uneven duty cycle. It can be enabled via register control, by setting register *0x457*, bit[1], to 1.

**Output Enable** (active-low, **OEB**). The output data drivers can be placed in a high-impedance state using the OEB function, accessed by setting register *0x457*, bit[3], to 1.

**External Reference Bias Resistor.** The SD7226 requires a 30kΩ, 1% resistor, between pin RBIAS and AVSS to set the default bias current for the ADC.

**Common-Mode Voltage.** The VCM pin outputs the analog input common-mode bias voltage (VCM). This pin provides a stable DC reference that can be used to set the ADC input common-mode level, either directly through a passive network or indirectly as the common-mode reference for a driving amplifier. In typical applications, the default common-mode voltage is suitable and requires no adjustment. The common-mode voltage level is programmable through register *0xf11*, bits [4:3].

**VREF.** The ADC supports both internal (default) and external voltage reference sources, selectable via the VREF pin.

The selection between internal and external reference voltage is via made register control: setting register *0x45d*, bit[5], to 1 selects the external reference.

Internal Reference Output. When the internal voltage reference is used, the VREF pin can function as a reference output. This output can be enabled by setting register *0x45d*, bits [14:13], to 0x3.

**Overrange Indicator.** This output bit indicates that the analog input sample exceeded the converter's full-scale input range. When the input signal exceeds the converter's positive (or negative) full-scale range, the ADC will clip, and the output data will saturate to the maximum (or minimum) digital code. In either case, the OR bit will be set, indicating that the input had been out of range during that sample.

The internal vref can be adjusted with *0x45d* bits [4:1]. Usually, the default value yields the optimal performance.

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## DSP

The DSP path, shown in Figure 27, includes functionality for processing the output data of two ADCs operated as an IQ pair or independently. All functions can be bypassed or disabled if not needed. The main blocks in the signal chain are input multiplexer, sample delay control, gain and offset correction, phase correction, digital down converter (DDC), and decimator. Refer to APP-NOTE 100 (*Understanding the DSP Features in Plural™ Family of ADCs*) for more details on this topic.

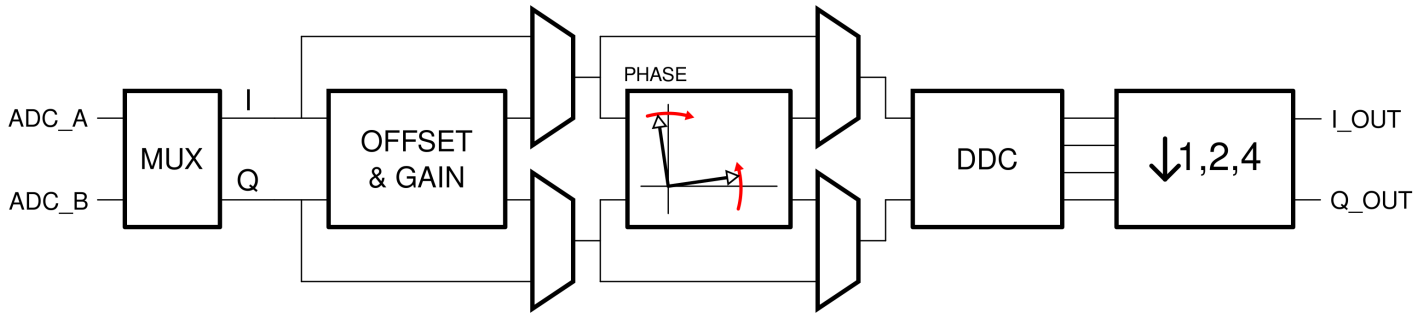



*Figure 27: SD7226 DSP Top-Level Diagram.*

## Input MUX

The DSP block has two inputs, both with a multiplexer for selecting the source signal. By default ADC B feeds the I (in-phase) input and ADC A the Q (quadrature) input. The MUX allows crossing the inputs, feeding the same ADC to both inputs, and using the average of the two ADC outputs as a source. The multiplexer is controlled by register *0xf09*.

## Delay, Offset, and Gain

Datapath configiration and sample delay control bits are in register *0x807*. Gain and offset are controlled by registers *0x809*, *0x80b*, *0x80d*, and *0x80f*.

## IQ Phase Correction

The block diagram of IQ phase correction block is shown in Figure 28. It can be used to add or subtract a small amount of I signal form the Q and vice versa to compensate for the phase error of the input signal. The correction range is about ±14 degrees and it is controlled by the *0x811* register. This simple scheme affects both the signal phase and gain and thus has to be used together with the gain correction block to minimize both the gain and phase mismatch.

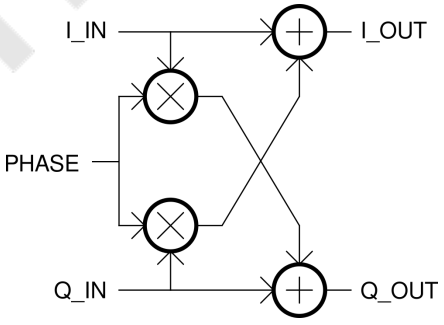



*Figure 28: SD7226 IQ phase correction.*

## Digital Down Converter (DDC)

Figure 29 shows the block diagram the DDC. It consists of two numerically controlled oscillators (NCOs), digital multipliers, adders, and multiplexers. It can be programmed for two modes of operation. Mode 1 provides two independent paths with a real input and a complex (IQ) output. Mode 2 is used when the two ADCs operate as an IQ pair. It accepts IQ input and

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supplies IQ output using NCO1 as local oscillator (LO).

The DDC is enabled in the datapath using a bit in register *0x807*. Its control bits are collected in register *0x813*, which has bits for enabling the I and Q paths and selecting the NCOs. The mode is selected using bits [14:13], value 0b01 selecting mode 1 and value 0b10 mode 2. In mode 2 the spectral inversion can be done by manipulating the signs of the adder inputs using bits [11:8], value 0b1101 is the default and value 0b0001 inverts the frequency axis.

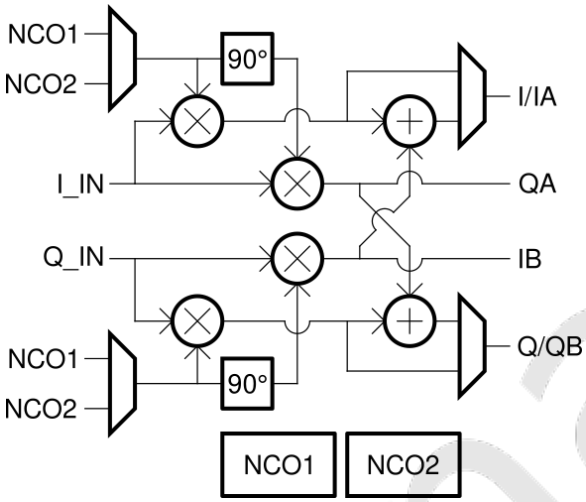



*Figure 29: SD7226 DDC.*

## NCO

The two NCO instances are implemented using a direct digital synthesizer (DDS). It can produce a LO frequency ranging from 0 to $F_s$ using a 48-bit frequency control word, which is calculated as $f/F_s*2^{48}$. To program the frequency, the DDC must be first enabled and a stable input clock must be present. Registers *0x815*, *0x817*, and *0x819* control NCO1 and registers *0x81b*, *0x81d*, and *0x81f* control NCO2. After the new frequency is written to the registers it has to be loaded by setting load bit high and then clearing it. NCO1 uses bit [0] and NCO2 bit [1] in register *0x82b* for this purpose.

## Decimator

The final block in the DSP signal chain is the decimator, which supports rates 2 and 4 or can be bypassed. It is based on two cascaded half-band FIR filters with the frequency response shown in Figure 30. The operation of the decimator is controlled by three registers *0x82f*, *0x82d*, and *0x805* that configure the decimator, set the rate, and DSP output clock divider.

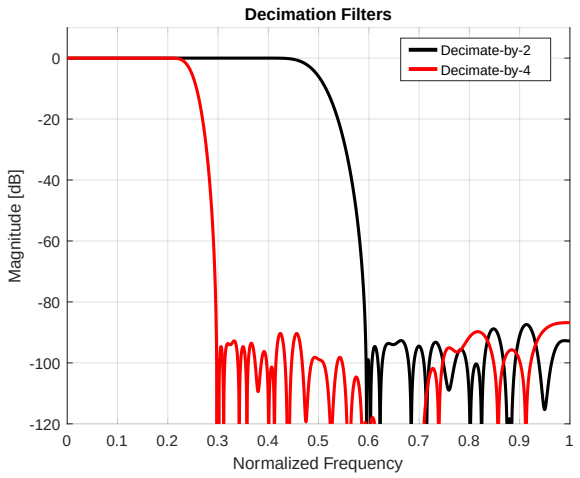



*Figure 30: SD7226 Decimator Frequency Response.*

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Table 19 provides the register values for each mode of operation.

*Table 19. Decimation Modes.*

| Mode | DSP Output Clock Rate | Reg *0x82f* | Reg *0x82d* | Reg *0x805* |
|---|---|---|---|---|
| Bypass | 1 | 0x0001 | 0x0000 | 0x0000 |
| Decimate by 2 | 1/2 | 0x0006 | 0x0010 | 0x0001 |
| Decimate by 4 | 1/4 | 0x0026 | 0x0020 | 0x0002 |
| DDC mode 1 & Decimate by 2 | 1 | 0x000E | 0x0010 | 0x0000 |
| DDC mode 1 & Decimate by 4 | 1/2 | 0x002E | 0x0020 | 0x0001 |
| Interleave & Decimate by 2 | 1/2 | 0x0016 | 0x0000 | 0x0000 |
| Interleave & Decimate by 4 | 1/4 | 0x0032 | 0x0010 | 0x0001 |

## Interleave and Decimate Mode

To operate the two ADC cores as an interleaved unit the input clock must be provided at twice the core sampling rate. The decimator is programmed per Table 19 and the ADCs are configured for interleaving using the following steps, which can only be performed after a stable input clock is present. Interleaving without decimation is not supported.

1. Write 0x0022 to register *0xf0f* to enable clock division by two.
2. Write 0x0000 to register *0xcc3* and 0x0002 to register *0xdc3* to set the clocks of the two ADC cores 180 degrees apart.
3. Toggle clock divider reset bit by first writing 0x0003 and then 0x0043 to register *0xf07*.
4. Program digital logic to work with 180 degree sampling clock offset by writing 0x000C to register *0xde7*.

In this mode the digital output data is present on both ADC A and ADC B output pins. The redundant output pins can be turned off to save power and reduce IO noise.

## Down Mixing Before Decimation

When decimating by two, the signals located in the upper half of the first Nyquist zone fall into the stop band of the decimation filter. Similarly the decimate by four mode passes signals up to $0.1F_s$ while blocking them from there on up to $0.9F_s$. The DDC block has a low power mode that uses a simple sequence of +1 and -1 multiplications that can mix the signal down with LO at $F_s/2$ or $F_s/4$ before decimation. This is enabled by the following steps, which can be exececuted only after a stable input clock is present.

1. Enable DDC in the datapath by clearing register *0x807* bit[9].
2. Write 0x59AB for LO of $F_s/2$ and 0x392B for LO at $F_s/4$ to register *0x813*.
3. Load the LO frequency by first writing 0x0001 and then 0x0000 to register *0x82b*.
4. Enable DDC low power mode by writing 0x59EB for LO of $F_s/2$ and 0x396B for LO of $F_s/4$ to register *0x813*.

An alternative and more versatile way to mix the signal down before decimation is to use the DDC. One significant difference is the digital output in IQ format.

In interleave and decimate mode, programming the part for mixing down with $F_s/4$ is the same as mixing down with $F_s/2$ above, as the definition of $F_s$ has changed by a factor of two. Setting the LO at $F_s/2$, which means multiplying every other sample by +1 and every other by -1, can be accomplished in two ways: in the board level by crossing the + and - of the analog input of one of the converters or using the following steps.

1. Enable DDC in the datapath by clearing register *0x807* bit [9].
2. Write 0x58AB to register *0x813*.
3. Set register *0x807* bit [0].
4. Load the LO frequency by first writing 0x0001 and then 0x0000 to register *0x82b*.
5. Clear register *0x807* bit [0].
6. Enable DDC low power mode by writing 0x58EB to register *0x813*.

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## Digital IOs

When the DSP is enabled the I-output is mapped to the ADC B output pins and the Q-output is mapped to the ADC A output pins. The rate of the digital output clock relative to the sampling clock (input clock after the clock divider) is specified in Table 19.

When the DDC is programmed to mode 1, it produces two IQ output signals each of which is interleaved into the corresponding ADC's output pins sending out the I and Q samples on alternate clock cycles. This output mode is enabled by writing 0b01 to register *0x4b5*, bits [10:9]. The ADC B overrange pin (not available in channel multiplexed mode, for the LVDS interface) can be programmed to serve as IQ indicator that toggles on every output clock cycle, by setting register *0x4cd*, bit [7]. A high value indicates a Q sample in A output and I sample in B output.

This mode suffers from a startup condition that instead of pairing sample Q(n) with sample I(n), sometimes Q(n) sample is paired with sample I(n-1) or I(n+1). Which of the two states is active can be detected with the following procedure and then handled properly in the receiver.

1. Program DSP to the desired mode of operation.
   a. If decimating by 2, program the DDS to frequency 0x200000000000
   b. if decimating by 4, program the DDS to frequency 0x100000000000
2. Load the frequency as described earlier (this creates a deterministic digital test pattern).
3. Set DDC output mux (register *0x813*, bits [14:13]) to 0b11 to enable the test pattern.
4. Observe the MSB of the received digital signal de-interleved into I and Q. Correct pairing shows the MSB of the Q-channel leading the I-channel by one sample. The correct state can be identified by delaying the MSB of Q and comparing it to the MSB of I. If any bit in the four-sample-long test pattern shows a mismatch, the samples are paired incorrectly. They should be realigned by applying a one-sample delay to Q of the B output and I of the A output in the receiver. The detection can be made with a simple XOR based circuit (see Listing 1).
5. Restore DDC output mux value to 0b01 and program the DDS(s) to the desired LO frequency.

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*Listing 1: The IQ-alignment SystemVerilog Module.*

```
module iq_align (
   input  logic clk,               // clock
   input  logic [15:0] da, db,     // 16-bit input data streams
   input  logic orb,               // Channel B overrange bit. Its second function identifies
                                   // whether the current data belongs to Q or I
   input  logic sel,               // Select between da and db, for alignment check
   input  logic align_now,         // Pulse to perform alingment. Must be minimum 4 clocks long
   input  logic capture_iq_flag,   // Pulse to capture orb
   input  logic reset_delay,       // Reset delay to 0
   output logic [15:0] da_i, da_q, // I/Q components of channel A
   output logic [15:0] db_i, db_q  // I/Q components of channel B
);
   // Synchronize locally generated I/Q indicator (iq_flag) to the one provided by
   // the chip on 'orb' pin.
   logic iq_flag;
   always_ff @(posedge clk)
      iq_flag <= ~(capture_iq_flag ? orb : iq_flag);

   // Split the input data into I and Q streams
   logic [15:0] da_i0, db_i0, da_q0, db_q0, da_i1, db_q1;
   always_ff @(posedge clk) begin
      if (iq_flag) begin
         da_i0 <= da; // High flag indicates I on channel A
         db_q0 <= db; // High flag indicates Q on channel B
      end else begin
         da_q0 <= da;
         db_i0 <= db;
         da_i1 <= da_i0;
         db_q1 <= db_q0;
      end
   end

   // Output data is launched on every other clock cycle, applying a delay to da_i and db_q
   // when misalignment is detected
   logic da_q_d_15, db_q_d_15, delay;
   always_ff @(posedge clk)
      if (~iq_flag) begin
         da_q <= da_q0;
         da_i <= delay ? da_i1 : da_i0;
         da_q_d_15 <= da_q[15];
         db_i <= db_i0;
         db_q <= delay ? db_q1 : db_q0;
         db_q_d_15 <= db_q[15];
      end

   // Compare the MSB of I and delayed Q to detect misalignment.
   // When the alingment is correct, the the signals are identical.
   wire misaligned = sel ? da_i[15]^da_q_d_15 : db_i[15]^db_q_d_15;

   // Enable delay, if misalignment is detected.
   always_ff @(posedge clk)
      if (reset_delay)
         delay <= 1'b0;
      else if (align_now & misaligned)
         delay <= 1'b1;
endmodule
```

Listing 1 shows an example SystemVerilog implementation of the alignment detection and correction logic. This implementation is provided as a reference and may require adjustments to meet specific customer requirements.

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## SERIAL PORT INTERFACE

The SD7226 uses a 3-wire Serial Port Interface (SPI) that gives the user flexibility to configure the converter for specific functions, depending on the application, through a register space provided inside the ADC. The interface signals are:

- SCLK: defines the bit rate at which serial data is driven onto, and sampled from, the bus;
- CSB: defines the boundaries of a basic data 'unit', comprised of multiple serial bits;
- SDIO: is the serial data IO wire;

The read and write cycles are described in the figure below. The address space is 13 bits long (A<12:0>) and the data is 16 bits wide (D<15:0>). The complete instruction cycle is 32-bits long. The falling edge of CSB combines with the rising edge of SCLK marks the start of the instruction cycle. On a write transaction, the target register is updated on the falling edge of SCLK.

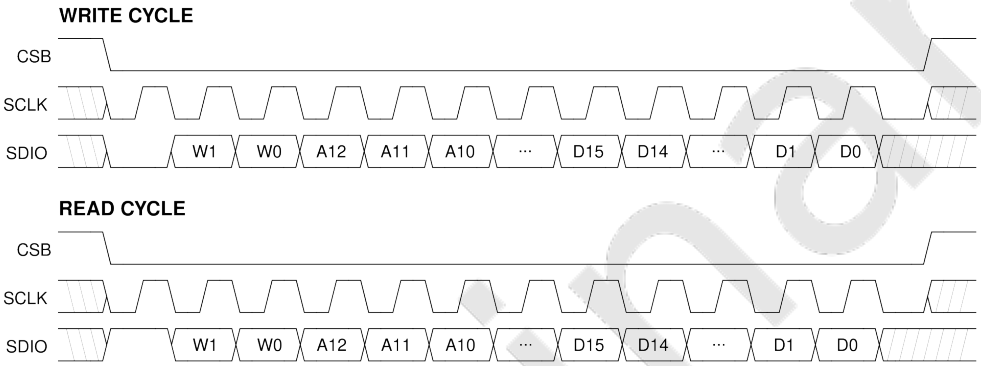



*Figure 31: 3-wire SPI timing.*

*Table 20. SPI Timing.*

| PARAMETER | TEMP | MIN | TYP | MAX | UNIT |
|---|---|---|---|---|---|
| Setup time between the data and the rising edge of SCLK | $t_{DS}$ | 2 | | | ns |
| Hold time between the data and the rising edge of SCLK | $t_{HD}$ | 2 | | | ns |
| Period of the SCLK | $t_{SCLK}$ | 40 | | | ns |
| Setup time between CSB and SCLK | $t_{S,CSB-SCLK}$ | 2 | | | ns |
| Hold time between CSB and SCLK | $t_{H,CSB-SCLK}$ | 2 | | | ns |
| Minimum period that SCLK should be in a logic high state | $t_{SCLK,high}$ | 10 | | | ns |
| Minimum period that SCLK should be in a logic high state | $t_{SCLK,low}$ | 10 | | | ns |
| Time from SCLK falling edge to SDIO switching from input to output | $t_{SDIO,EN}$ | 10 | | | ns |
| Time from SCLK falling edge to SDIO switching from output to input | $t_{SDIO,DIS}$ | 10 | | | ns |

Two bits, W1 and W0, determine how many bytes of data that can be transferred in the same write cycle (see Table 21). If more than 16 bits (2 Bytes) of data are being transferred the address is increased sequentially.

*Table 21. SPI Word Length.*

| [W1,W0] | Data length |
|---|---|
| 00 | Not supported |
| 01 | Two bytes of data can be transferred |
| 10 | Not supported |
| 11 | Four bytes of data can be transferred |

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The SPI pins should not be active when the full dynamic performance of the ADC is required. Noise from SCLK, CSB and the data transactions can degrade ADC performance.

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# OUTPUT MODE

## Data Scrambler

Interference originating from the digital outputs of the ADC can be difficult to eliminate entirely. Such interference may result from capacitive or inductive coupling mechanisms, or from shared impedance paths in the ground plane. Even minimal coupling coefficients can introduce deterministic spurious tones into the ADC's output frequency spectrum. To mitigate this, digital output scrambling techniques can be employed prior to off-chip transmission. By randomizing the bit patterns, the spectral energy of these spurs is dispersed, effectively reducing their peak amplitudes and minimizing their impact on signal integrity.

The SD7226 can apply an exclusive-OR logic operation between the LSB and all other data output bits, while the LSB, overflow and clock outputs are not affected.

When this function is used, the receiver must apply the same function to unscramble the received data.

The data scrambler is enabled by setting control register *0x4b5*, bit[13], to 1.

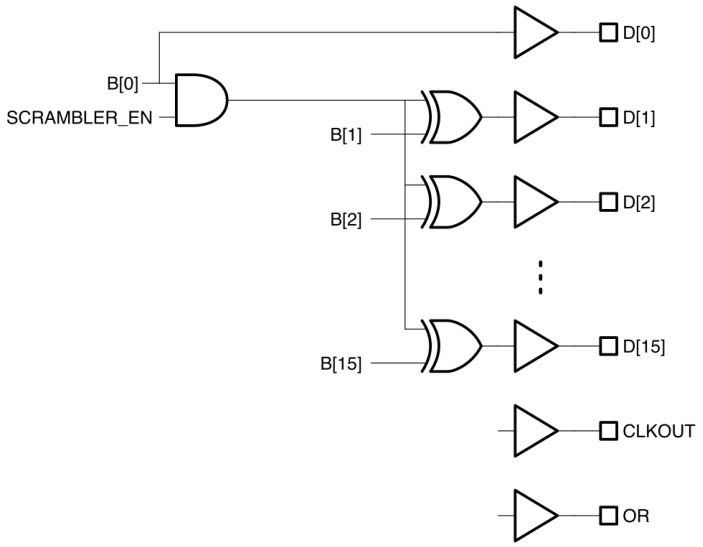



*Figure 32: SD7226 Data Scrambler.*

## Alternate Bit Polarity

The alternate bit polarity is particularly effective to suppress digital feedback and minimize noise coupling on the PCB when the ADC input signal is near mid-scale and of very small amplitude. In this case, the digital output tends to toggle between patterns dominated by either logic high or logic low states. This synchronized switching of multiple bits can induce significant transient currents in the ground plane, leading to increased digital noise. This mode, when activated, inverts all odd-numbered data output bits prior to the output buffer stage while even-numbered bits, along with the overflow and clock output, remain unaffected thus ensuring that approximately half of the output bits transition are high while the other half transition are low. This balanced switching behavior helps cancel out opposing current flows in the ground return path, thereby reducing overall ground noise. At the receiving end, the original data can be reconstructed by inverting the same odd-numbered bits. This mode operates independently of the digital output randomization feature, both functions can be enabled or disabled separately. The Alternate Bit Polarity mode is configured via serial programming of control register *0x4b5*, bit[14].

## Output Test Modes

The output test options are described in Table 22 and are selected via SPI programming at register *0x4b5*, bits [5:2].

When an output test mode is enabled, the analog section of the ADC is disconnected from the digital back-end blocks and the test pattern is run through the output formatting block. These tests require an active input clock.

There are two pseudo-random number generators available, PN23 and PN9. The PN23 generator ($X^{23}+X^{18}+1$), selected by register *0x4b5*, bits [5:2], set to 0x5, can be reset by setting register *0x4b5*, bit [12], low; while the PN9 generator ($X^{9}+X^{5}+1$), selected by register *0x4b5* set to 0x6, can be reset by setting register *0x4b5*, bit [11], low.

*Table 22. Output Test Modes.*

| Mode | Function |
|---|---|
| 0 | Pass-Through |
| 1 | Midscale |
| 2 | +FS |
| 3 | -FS |
| 4 | Checkerboard |
| 5 | PN23 |
| 6 | PN9 |
| 7 | 1/0 word toggle |
| 8 | User input |
| 9 | 1/0 bit toggle |
| 10 | Unused |
| 11 | 1-bit high |
| 12 | Mixed frequency |
| 13 | Unused |
| 14 | Unused |
| 15 | Ramp |

Pass-through and test patterns 1, 2, 3, 5 and 6 are subject to output formatting, while the other test modes are not. Test patterns 1 to 8 and 15 can be applied to either one channel or both channels using *0x4b5*, bit[7:6]. Patterns 4 and 7 can be toggled between the pattern and its inverse while test pattern 8 can be selected using registers *0x4bb*, *0x4bd*, *0x4bf* and *0x4c1*.

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## CONTROL REGISTERS

This section describes the most commonly used control registers. For a description of the complete register map, refer to Appendix A: Register Map.

Each register is presented in the following format:

| **Address: <HEX value>** | | **RW** or **RO** | **Default: <HEX value>** |
|---|---|---|---|
| bit field | default value | field description. | |

### Notes

- **RW** = Read/Write
- **RO** = Read-Only
- For read-only registers, the default value is omitted.
- Registers not listed in these tables should not be written.

When updating control register values, always use a read-modify-write procedure. Some registers include reserved bit fields for internal engineering purposes. This approach ensures that these reserved bits remain unchanged and prevents unintended modifications.

**CHIP TOP**:

| **Address: 0x457** | | **RW** | **Default: 0x0015** |
|---|---|---|---|
| [0] | 0x1 | Duty Cycle Stabilizer enable (see Duty-Cycle Stabilizer).<br>0: Disable<br>1: Enable | |
| [1] | 0x0 | Duty Cycle Stabilizer on/off in SPI mode (see Duty-Cycle Stabilizer).<br>0: Off<br>1: On | |
| [2] | 0x1 | External output enable (OE/OEB) control (see Output Enable).<br>0: Output Enabled<br>1: Output depends on pin OE/OEB | |
| [3] | 0x0 | Output disable (see Output Enable).<br>0: Output depends on OE/OEB control and OE/OEB pin<br>1: Disabled | |
| [6:4] | 0x1 | Reserved. | |
| [7] | 0x0 | Software power down.<br>0: Active State<br>1: Power Down | |
| [8] | 0x0 | Software stand-by mode.<br>0: Active State<br>1: Stand-by Mode | |

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| Address: 0x463 | | RW | Default: 0x0019 |
|---|---|---|---|
| [0] | 0x1 | Enable clock receiver.<br>0: Disable<br>1: Enable | |
| [4:1] | 0xc | Reserved. | |
| [7:5] | 0x0 | Clock RX divider control.<br>0: Bypass<br>1: Divide by 2<br>2: Divide by 3<br>3: Divide by 4<br>4: Divide by 5<br>5: Divide by 6<br>6: Divide by 7<br>7: Divide by 8 | |
| [8] | 0x0 | Enable on-chip 100Ohm termination.<br>0: Disable<br>1: Enable | |
| [9] | 0x0 | Clock polarity control when DCS is enabled.<br>0: Disable<br>1: Invert | |
| [10] | 0x0 | Enable SYNC pin synchronization.<br>0: Disable<br>1: Enable | |

| Address: 0x465 | | RW | Default: 0x0004 |
|---|---|---|---|
| [2:0] | 0x4 | Output data interleaver.<br>0: pass through, no interleaving<br>1: pass through, channels swapped, no interleaving<br>2: N/A<br>3: N/A<br>4: parallel ADC_A/ADC_B interleaving<br>5: parallel ADC_B/ADC_A interleaving<br>6: even-odd interleaving<br>7: odd-even interleaving | |

| Address: 0x473 | | RW | Default: 0x0000 |
|---|---|---|---|
| [4:0] | 0x0 | Reserved. | |
| [8:5] | 0x0 | Output data delay in parallel CMOS mode.<br>The output data delay (250ps steps) with respect to the clock | |
| [9] | 0x0 | Launching output clock polarity in parallel CMOS mode.<br>0: Not inverted<br>1: Inverted | |
| [13:10] | 0x0 | Output clock delay.<br>The output clock delay is increased by 250ps * (register value) with respect to the data | |
| [14] | 0x0 | Output clock polarity.<br>0: Not inverted<br>1: Inverted | |

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| Address: 0x4b5 | | RW | Default: 0x18c0 |
|---|---|---|---|
| [1:0] | 0x0 | Output format.<br>0: Offset binary<br>1: Two's complement<br>2: Gray code<br>3: Offset binary | |
| [5:2] | 0x0 | Output test mode select (see Table 22). | |
| [6] | 0x1 | Output test channel CH. A<br>0: Disable<br>1: Enable | |
| [7] | 0x1 | Output test channel CH. B<br>0: Disable<br>1: Enable | |
| [8] | 0x0 | Output test toggle mode. Toggle between user test pattern 0 and 1. | |
| [10:9] | 0x0 | Reserved. | |
| [11] | 0x1 | PN9 generator reset (active low). | |
| [12] | 0x1 | PN23 generator reset (active low). | |
| [13] | 0x0 | Enable the output data scrambler.<br>0: Disable<br>1: Enable | |
| [14] | 0x0 | Enable alternate bit polarity switch.<br>0: Disable<br>1: Enable | |

**ADC A**:

| Address: 0xdc1 | | RW | Default: 0xbfff |
|---|---|---|---|
| [3:0] | 0xf | Reserved. | |
| [4] | 0x1 | Calibration soft reset (active low). | |
| [15:5] | 0x5f9 | Reserved. | |

See Table 18 for valid configurations for the following registers:

| Address: 0xdcd | | RW | Default: 0x1483 |
|---|---|---|---|
| [2:0] | 0x3 | Stg1 calibration convergence speed in foreground mode (see Table 18). | |
| [11:3] | 0x90 | Reserved. | |
| [13:12] | 0x1 | Stg1 calibration randomization sequence in foreground mode (see Table 18). | |

| Address: 0xdd5 | | RW | Default: 0x1483 |
|---|---|---|---|
| [2:0] | 0x3 | Stage 2 calibration convergence speed in foreground mode (see Table 18). | |
| [11:3] | 0x90 | Reserved. | |
| [13:12] | 0x1 | Stage 2 calibration randomization sequence in foreground mode (see Table 18). | |

| Address: 0xddd | | RW | Default: 0x16c3 |
|---|---|---|---|
| [2:0] | 0x3 | Stage 3 calibration convergence speed in foreground mode (see Table 18). | |
| [11:3] | 0xd8 | Reserved. | |
| [13:12] | 0x1 | Stage 3 calibration randomization sequence in foreground mode (see Table 18). | |

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| Address: 0xde1 | | RW | Default: 0xbbe8 |
|---|---|---|---|
| [7:0] | 0xe8 | Time allocated for calibration (see Table 18). | |
| [8] | 0x1 | Enable calibration (see Table 18). | |
| [9] | 0x1 | Enable foreground calibration (see Table 18). | |
| [10] | 0x0 | Reserved. | |
| [11] | 0x1 | Enable stg1 calibration (see Table 18). | |
| [12] | 0x1 | Enable stage 2 calibration (see Table 18). | |
| [13] | 0x1 | Enable stage 3 calibration (see Table 18). | |
| [14] | 0x0 | Background calibration parallel mode (see Table 18). | |
| [15] | 0x1 | Disable background calibration mode for all stages (see Table 18). | |

**ADC B**:

| Address: 0xcc1 | | RW | Default: 0xbfff |
|---|---|---|---|
| [3:0] | 0xf | Reserved. | |
| [4] | 0x1 | Calibration soft reset (active low). | |
| [15:5] | 0x5f9 | Reserved. | |

See Table 18 for valid configurations for the following registers:

| Address: 0xccd | | RW | Default: 0x1483 |
|---|---|---|---|
| [2:0] | 0x3 | Stg1 calibration convergence speed in foreground mode (see Table 18). | |
| [11:3] | 0x90 | Reserved. | |
| [13:12] | 0x1 | Stg1 calibration randomization sequence in foreground mode (see Table 18). | |

| Address: 0xcd5 | | RW | Default: 0x1483 |
|---|---|---|---|
| [2:0] | 0x3 | Stage 2 calibration convergence speed in foreground mode (see Table 18). | |
| [11:3] | 0x90 | Reserved. | |
| [13:12] | 0x1 | Stage 2 calibration randomization sequence in foreground mode (see Table 18). | |

| Address: 0xcdd | | RW | Default: 0x16c3 |
|---|---|---|---|
| [2:0] | 0x3 | Stage 3 calibration convergence speed in foreground mode (see Table 18). | |
| [11:3] | 0xd8 | Reserved. | |
| [13:12] | 0x1 | Stage 3 calibration randomization sequence in foreground mode (see Table 18). | |

| Address: 0xce1 | | RW | Default: 0xbbe8 |
|---|---|---|---|
| [7:0] | 0xe8 | Time allocated for calibration (see Table 18). | |
| [8] | 0x1 | Enable calibration (see Table 18). | |
| [9] | 0x1 | Enable foreground calibration (see Table 18). | |
| [10] | 0x0 | Reserved. | |
| [11] | 0x1 | Enable stg1 calibration (see Table 18). | |
| [12] | 0x1 | Enable stage 2 calibration (see Table 18). | |
| [13] | 0x1 | Enable stage 3 calibration (see Table 18). | |
| [14] | 0x0 | Background calibration parallel mode (see Table 18). | |
| [15] | 0x1 | Disable background calibration mode for all stages (see Table 18). | |

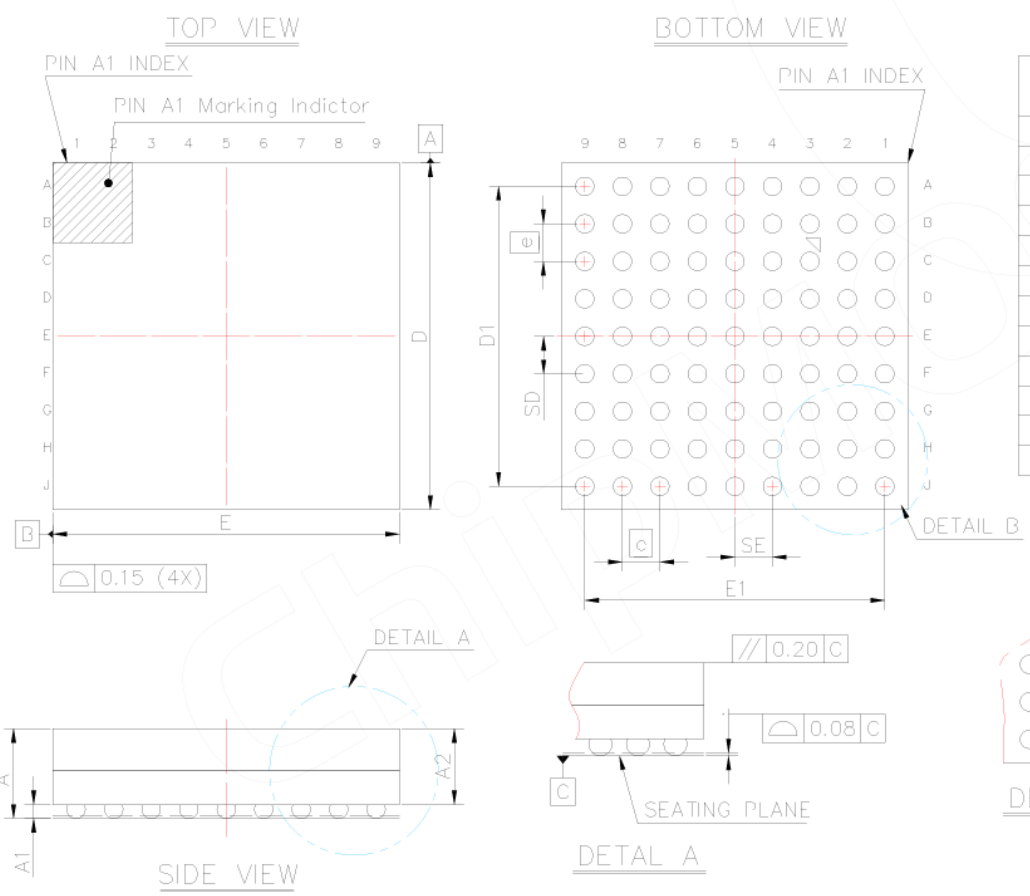
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## ADC BRING-UP

The SD7226 does not require a specific power-up sequence; however, it is recommended to apply the DRVDD first followed by the AVDD.

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## PACKAGE DRAWING




| SYM. | DIMENSION(mm) | | |
|---|---|---|---|
| | MIN. | NOM. | MAX. |
| A | 0.90 | 0.96 | 1.10 |
| A1 | 0.10 | 0.15 | 0.20 |
| A2 | - | 0.81 | - |
| b | 0.15 | 0.20 | 0.25 |
| D | 3.60 | 3.70 | 3.80 |
| D1 | 3.20 BSC | | |
| E | 3.60 | 3.70 | 3.80 |
| E1 | 3.20 BSC | | |
| SD | 0.40 BSC | | |
| SE | 0.40 BSC | | |
| e | - | 0.40 | - |
| Ball size pre-reflow | | | 0.20 |

*Figure 33: Package Dimensions.*

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# APPENDIX A: Register Map

This section provides a comprehensive description of the complete register map, detailing all available control and status registers within the device.

## 0x400 - CHIP_TOP

| Address: 0x455 | | RW | Default: 0x0001 |
|---|---|---|---|
| [0] | 0x1 | Soft reset (active low). | |

| Address: 0x457 | | RW | Default: 0x0015 |
|---|---|---|---|
| [0] | 0x1 | Duty Cycle Stabilizer enable (see Duty-Cycle Stabilizer).<br>0: Disable<br>1: Enable | |
| [1] | 0x0 | Duty Cycle Stabilizer on/off in SPI mode (see Duty-Cycle Stabilizer).<br>0: Off<br>1: On | |
| [2] | 0x1 | External output enable (OE/OEB) control (see Output Enable).<br>0: Output Enabled<br>1: Output depends on pin OE/OEB | |
| [3] | 0x0 | Output disable (see Output Enable).<br>0: Output depends on OE/OEB control and OE/OEB pin<br>1: Disabled | |
| [6:4] | 0x1 | Reserved. | |
| [7] | 0x0 | Software power down.<br>0: Active State<br>1: Power Down | |
| [8] | 0x0 | Software stand-by mode.<br>0: Active State<br>1: Stand-by Mode | |

| Address: 0x45d | | RW | Default: 0x0095 |
|---|---|---|---|
| [0] | 0x1 | ADC reference voltage enable.<br>0: Disabled<br>1: Enabled | |
| [4:1] | 0xa | ADC reference voltage value.<br>0: Lowest<br>...<br>15: Highest | |
| [5] | 0x0 | Set reference.<br>0: Internal<br>1: External from VREF pin | |
| [11:6] | 0x2 | Reserved. | |
| [12] | 0x0 | Enable selection between Internal and External VREF selection via bit 5.<br>0: Bit [5] selection disabled<br>1: Bit [5] selection enabled | |
| [13] | 0x0 | Enable selection of VREF out via bit 14.<br>0: Bit [14] selection disabled<br>1: Bit [14] selection enabled | |
| [14] | 0x0 | Enable VREF pin as reference voltage output.<br>0: Disabled<br>1: Enabled | |

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| Address: 0x463 | | RW | Default: 0x0019 |
|---|---|---|---|
| [0] | 0x1 | Enable clock receiver.<br>0: Disable<br>1: Enable | |
| [4:1] | 0xc | Reserved. | |
| [7:5] | 0x0 | Clock RX divider control.<br>0: Bypass<br>1: Divide by 2<br>2: Divide by 3<br>3: Divide by 4<br>4: Divide by 5<br>5: Divide by 6<br>6: Divide by 7<br>7: Divide by 8 | |
| [8] | 0x0 | Enable on-chip 100Ohm termination.<br>0: Disable<br>1: Enable | |
| [9] | 0x0 | Clock polarity control when DCS is enabled.<br>0: Disable<br>1: Invert | |
| [10] | 0x0 | Enable SYNC pin synchronization.<br>0: Disable<br>1: Enable | |

| Address: 0x465 | | RW | Default: 0x0004 |
|---|---|---|---|
| [2:0] | 0x4 | Output data interleaver.<br>0: pass through, no interleaving<br>1: pass through, channels swapped, no interleaving<br>2: N/A<br>3: N/A<br>4: parallel ADC_A/ADC_B interleaving<br>5: parallel ADC_B/ADC_A interleaving<br>6: even-odd interleaving<br>7: odd-even interleaving | |

| Address: 0x467 | | RW | Default: 0x2851 |
|---|---|---|---|
| [0] | 0x1 | LVDS enable.<br>0: Disable<br>1: Enable | |
| [1] | 0x0 | LVDS reduced swing mode.<br>0: Disable<br>1: Enable | |
| [10:2] | 0x14 | Reserved. | |
| [13:11] | 0x5 | LVDS amplitude.<br>0: Smallest amplitude<br>0: ...<br>7: Largest amplitude | |
| [15:14] | 0x0 | Reserved. | |

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| Address: 0x473 | | RW | Default: 0x0000 |
|---|---|---|---|
| [4:0] | 0x0 | Reserved. | |
| [8:5] | 0x0 | Output data delay in parallel CMOS mode.<br>The output data delay (250ps steps) with respect to the clock | |
| [9] | 0x0 | Launching output clock polarity in parallel CMOS mode.<br>0: Not inverted<br>1: Inverted | |
| [13:10] | 0x0 | Output clock delay.<br>The output clock delay is increased by 250ps * (register value) with respect to the data | |
| [14] | 0x0 | Output clock polarity.<br>0: Not inverted<br>1: Inverted | |

| Address: 0x475 | | RW | Default: 0x0001 |
|---|---|---|---|
| [0] | 0x1 | Enable VCM.<br>0: Disable<br>1: Enable | |

| Address: 0x477 | | RW | Default: 0x0000 |
|---|---|---|---|
| [1:0] | 0x0 | Sample delay for ADC A data | |
| [3:2] | 0x0 | Sample delay for ADC B data | |

| Address: 0x4b5 | | RW | Default: 0x18c0 |
|---|---|---|---|
| [1:0] | 0x0 | Output format.<br>0: Offset binary<br>1: Two's complement<br>2: Gray code<br>3: Offset binary | |
| [5:2] | 0x0 | Output test mode select (see Table 22). | |
| [6] | 0x1 | Output test channel CH. A<br>0: Disable<br>1: Enable | |
| [7] | 0x1 | Output test channel CH. B<br>0: Disable<br>1: Enable | |
| [8] | 0x0 | Output test toggle mode. Toggle between user test pattern 0 and 1. | |
| [10:9] | 0x0 | Reserved. | |
| [11] | 0x1 | PN9 generator reset (active low). | |
| [12] | 0x1 | PN23 generator reset (active low). | |
| [13] | 0x0 | Enable the output data scrambler.<br>0: Disable<br>1: Enable | |
| [14] | 0x0 | Enable alternate bit polarity switch.<br>0: Disable<br>1: Enable | |

| Address: 0x4b7 | | RW | Default: 0x0092 |
|---|---|---|---|
| [15:0] | 0x0092 | PN9 initial seed. | |

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| Address: 0x4b9 | | RW | Default: 0x3aff |
|---|---|---|---|
| [15:0] | 0x3aff | PN23 initial seed. | |

| Address: 0x4bb | | RW | Default: 0x0000 |
|---|---|---|---|
| [15:0] | 0x0000 | User test pattern 0 (CH. A). | |

| Address: 0x4bd | | RW | Default: 0x0000 |
|---|---|---|---|
| [15:0] | 0x0000 | User test pattern 1 (CH. A). | |

| Address: 0x4bf | | RW | Default: 0x0000 |
|---|---|---|---|
| [15:0] | 0x0000 | User test pattern 0 (CH. B). | |

| Address: 0x4c1 | | RW | Default: 0x0000 |
|---|---|---|---|
| [15:0] | 0x0000 | User test pattern 1 (CH. B). | |

| Address: 0x4c9 | | RW | Default: 0x0000 |
|---|---|---|---|
| [0] | 0x0 | Disable auto-recalibration.<br>0: Disable<br>1: Enable | |
| [15:1] | 0x0 | Reserved. | |

| Address: 0x4cd | | RW | Default: 0x0000 |
|---|---|---|---|
| [6:0] | 0x0 | Reserved. | |
| [7] | 0x0 | Enable IQ indicator (Digital IOs).<br>0: Disable<br>1: Enable | |
| [15:8] | 0x0 | Reserved. | |

| Address: 0x4cf | | RW | Default: 0x0035 |
|---|---|---|---|
| [14:0] | 0x35 | Reserved. | |
| [15] | 0x0 | Disable production test mode.<br>0: Test mode enabled<br>1: Test mode disabled | |

| Address: 0x4d3 | | RW | Default: 0x0007 |
|---|---|---|---|
| [3:0] | 0x7 | LVDS common mode voltage. | |
| [7:4] | 0x0 | Reserved. | |

| Address: 0x4fd | RO | Default: N/A |
|---|---|---|
| [2:0] | Chip revision. | |
| [6:3] | Label ID. | |
| [8:7] | Reserved. | |
| [10:9] | Resolution ID for device<br>0: 10-bit<br>1: 12-bit<br>2: 14-bit<br>3: 16-bit | |
| [13:11] | Speed ID for device. | |

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| Address: 0x4ff | RO | Default: N/A |
|---|---|---|
| [0] | Reserved. | |
| [1] | Analog Supply Ready.<br>0: Not ready<br>1: Ready | |
| [2] | Digital Supply Ready. | |
| [3] | I/O Supply Ready.<br>0: Not ready<br>1: Ready | |
| [4] | Reserved. | |
| [5] | SPI enabled.<br>0: Not enabled<br>1: Enabled | |
| [6] | Power down pin status.<br>0: Inactive<br>1: Active | |
| [9:7] | Reserved. | |
| [10] | SYNC pin status. | |

## 0xF00 - ADC_DUAL

| Address: 0x809 | | RW | Default: 0x1800 |
|---|---|---|---|
| [11:0] | 0x800 | A/Q channel gain.<br>0x000: 0.0<br>0x800: 1.0<br>0xfff: 2.0 | |
| [12] | 0x1 | Enable A/Q path gain and offset. | |

| Address: 0x80b | | RW | Default: 0x1800 |
|---|---|---|---|
| [11:0] | 0x800 | B/I channel gain.<br>0x000: 0.0<br>0x800: 1.0<br>0xfff: 2.0 | |
| [12] | 0x1 | Enable B/I path gain and offset. | |

| Address: 0x80d | | RW | Default: 0x0000 |
|---|---|---|---|
| [15:0] | 0x0000 | A/Q channel offset. Signed 16-bit two's complement number. | |

| Address: 0x80f | | RW | Default: 0x0000 |
|---|---|---|---|
| [15:0] | 0x0000 | B/I channel offset. Signed 16-bit two's complement number. | |

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| Address: 0xf03 | | RW | Default: 0x07f9 |
|---|---|---|---|
| [0] | 0x1 | Enable top level bias.<br>0: Disable<br>1: Enable | |
| [2:1] | 0x0 | Reserved. | |
| [3] | 0x1 | ADC_B enable.<br>0: Disable<br>1: Enable | |
| [4] | 0x1 | ADC_A enable.<br>0: Disable<br>1: Enable | |
| [10:5] | 0x3f | Mask for enable pin. | |

| Address: 0xf05 | | RW | Default: 0x0030 |
|---|---|---|---|
| [5:0] | 0x30 | Mask for stand-by pin. | |
| [6] | 0x0 | Stand-by ADC_B.<br>0: Disable<br>1: Enable | |
| [7] | 0x0 | Stand-by ADC_A.<br>0: Disable<br>1: Enable | |

| Address: 0xf07 | | RW | Default: 0x0043 |
|---|---|---|---|
| [0] | 0x1 | Digital clock enable.<br>0: Disable<br>1: Enable | |
| [1] | 0x1 | Analog clock enable.<br>0: Disable<br>1: Enable | |
| [2] | 0x0 | Reserved. | |
| [3] | 0x0 | Clock source select.<br>0: ADC_B<br>1: ADC_A | |
| [5:4] | 0x0 | Reserved. | |
| [6] | 0x1 | ADC clock generator reset (active low). | |
| [7] | 0x0 | Reserved. | |

| Address: 0xf09 | | RW | Default: 0x0040 |
|---|---|---|---|
| [2:0] | 0x0 | Channel B source.<br>000: ADC B (default)<br>011: ADC A<br>111: (ADC A + ADC B)/2 | |
| [5:3] | 0x0 | Channel A source.<br>000: ADC A (default)<br>011: ADC B<br>111: 0 | |
| [6] | 0x1 | Reserved. | |

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| Address: 0xf0f | | RW | Default: 0x0028 |
|---|---|---|---|
| [1:0] | 0x0 | Clock divider (see Interleave and Decimate Mode.<br>0: Bypass<br>1: Divide-by-2<br>2: N/A<br>3: N/A | |
| [5:2] | 0xa | Reserved. | |

| Address: 0xf11 | | RW | Default: 0x0003 |
|---|---|---|---|
| [2:0] | 0x3 | Reserved. | |
| [4:3] | 0x0 | VCM voltage control.<br>0: Lowest common-mode voltage<br>...<br>3: Highest common-mode voltage | |

## 0xDC0 - ADC A

| Address: 0xdc1 | | RW | Default: 0xbfff |
|---|---|---|---|
| [3:0] | 0xf | Reserved. | |
| [4] | 0x1 | Calibration soft reset (active low). | |
| [15:5] | 0x5f9 | Reserved. | |

| Address: 0xdc3 | | RW | Default: 0x0000 |
|---|---|---|---|
| [3:0] | 0x0 | Clock phase. | |
| [14:4] | 0x0 | Reserved. | |

| Address: 0xdc9 | | RW | Default: 0x06a5 |
|---|---|---|---|
| [8:0] | 0xa5 | Reserved. | |
| [9] | 0x1 | Stg1 enable background mode. | |
| [11:10] | 0x1 | Reserved. | |

| Address: 0xdcd | | RW | Default: 0x1483 |
|---|---|---|---|
| [2:0] | 0x3 | Stg1 calibration convergence speed in foreground mode (see Table 18). | |
| [11:3] | 0x90 | Reserved. | |
| [13:12] | 0x1 | Stg1 calibration randomization sequence in foreground mode (see Table 18). | |

| Address: 0xdcf | | RW | Default: 0x0d85 |
|---|---|---|---|
| [2:0] | 0x5 | Stg1 calibration convergence speed in background mode (see Table 18). | |
| [11:3] | 0x1b0 | Reserved. | |

| Address: 0xdd1 | | RW | Default: 0x02a5 |
|---|---|---|---|
| [0] | 0x1 | Stage 2 gain error calibration enable. | |
| [1] | 0x0 | Stage 2 gain error calibration only. | |
| [8:2] | 0x29 | Reserved. | |
| [9] | 0x1 | Stage 2 enable background mode. | |
| [11:10] | 0x0 | Reserved. | |

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| Address: 0xdd3 | | RW | Default: 0x0bc0 |
|---|---|---|---|
| [0] | 0x0 | Stage 2 force calibration data. | |
| [3:1] | 0x0 | Stage 2 calibration data register. | |
| [6:4] | 0x4 | Stage 2 bg calibration data register. | |
| [9:7] | 0x7 | Stage 2 fg calibration data register. | |
| [11:10] | 0x2 | Stage 2 calibration mode. | |
| [14:12] | 0x0 | Stage 2 capacitor select register. | |
| [15] | 0x0 | Stage 2 cal data magnitude. | |

| Address: 0xdd5 | | RW | Default: 0x1483 |
|---|---|---|---|
| [2:0] | 0x3 | Stage 2 calibration convergence speed in foreground mode (see Table 18). | |
| [11:3] | 0x90 | Reserved. | |
| [13:12] | 0x1 | Stage 2 calibration randomization sequence in foreground mode (see Table 18). | |

| Address: 0xdd7 | | RW | Default: 0x0001 |
|---|---|---|---|
| [2:0] | 0x1 | Stage 2 calibration convergence speed in background mode (see Table 18). | |
| [11:3] | 0x0 | Reserved. | |

| Address: 0xddd | | RW | Default: 0x16c3 |
|---|---|---|---|
| [2:0] | 0x3 | Stage 3 calibration convergence speed in foreground mode (see Table 18). | |
| [11:3] | 0xd8 | Reserved. | |
| [13:12] | 0x1 | Stage 3 calibration randomization sequence in foreground mode (see Table 18). | |

| Address: 0xde1 | | RW | Default: 0xbbe8 |
|---|---|---|---|
| [7:0] | 0xe8 | Time allocated for calibration (see Table 18). | |
| [8] | 0x1 | Enable calibration (see Table 18). | |
| [9] | 0x1 | Enable foreground calibration (see Table 18). | |
| [10] | 0x0 | Reserved. | |
| [11] | 0x1 | Enable stg1 calibration (see Table 18). | |
| [12] | 0x1 | Enable stage 2 calibration (see Table 18). | |
| [13] | 0x1 | Enable stage 3 calibration (see Table 18). | |
| [14] | 0x0 | Background calibration parallel mode (see Table 18). | |
| [15] | 0x1 | Disable background calibration mode for all stages (see Table 18). | |

| Address: 0xde3 | | RW | Default: 0x0008 |
|---|---|---|---|
| [2:0] | 0x0 | Reserved. | |
| [3] | 0x1 | Clipping control to reserve some headroom for background calibration signals (active low). | |

| Address: 0xde7 | | RW | Default: 0x0004 |
|---|---|---|---|
| [2:0] | 0x4 | Reserved. | |
| [4:3] | 0x0 | Output data timing. (see Interleave and Decimate Mode) | |

| Address: 0xdeb | | RW | Default: 0x0000 |
|---|---|---|---|
| [0] | 0x0 | Calibration status.<br>0: Not done<br>1: Done | |
| [7:1] | 0x0 | Reserved. | |

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| Address: 0xdef | RO | Default: N/A |
|---|---|---|
| [0] | Calibration complete. | |
| [15:1] | Reserved. | |

## 0xCC1 - ADC B

| Address: 0xcc1 | RW | Default: 0xbfff |
|---|---|---|
| [3:0] | 0xf | Reserved. |
| [4] | 0x1 | Calibration soft reset (active low). |
| [15:5] | 0x5f9 | Reserved. |

| Address: 0xcc3 | RW | Default: 0x0000 |
|---|---|---|
| [3:0] | 0x0 | Clock phase. (see Interleave and Decimate Mode) |
| [14:4] | 0x0 | Reserved. |

| Address: 0xcc9 | RW | Default: 0x06a5 |
|---|---|---|
| [8:0] | 0xa5 | Reserved. |
| [9] | 0x1 | Stg1 enable background mode. |
| [11:10] | 0x1 | Reserved. |

| Address: 0xccd | RW | Default: 0x1483 |
|---|---|---|
| [2:0] | 0x3 | Stg1 calibration convergence speed in foreground mode (see Table 18). |
| [11:3] | 0x90 | Reserved. |
| [13:12] | 0x1 | Stg1 calibration randomization sequence in foreground mode (see Table 18). |

| Address: 0xccf | RW | Default: 0x0d85 |
|---|---|---|
| [2:0] | 0x5 | Stg1 calibration convergence speed in background mode (see Table 18). |
| [11:3] | 0x1b0 | Reserved. |

| Address: 0xcd1 | RW | Default: 0x02a5 |
|---|---|---|
| [0] | 0x1 | Stage 2 gain error calibration enable. |
| [1] | 0x0 | Stage 2 gain error calibration only. |
| [8:2] | 0x29 | Reserved. |
| [9] | 0x1 | Stage 2 enable background mode. |
| [11:10] | 0x0 | Reserved. |

| Address: 0xcd5 | RW | Default: 0x1483 |
|---|---|---|
| [2:0] | 0x3 | Stage 2 calibration convergence speed in foreground mode (see Table 18). |
| [11:3] | 0x90 | Reserved. |
| [13:12] | 0x1 | Stage 2 calibration randomization sequence in foreground mode (see Table 18). |

| Address: 0xcd7 | RW | Default: 0x0001 |
|---|---|---|
| [2:0] | 0x1 | Stage 2 calibration convergence speed in background mode (see Table 18). |
| [11:3] | 0x0 | Reserved. |

| Address: 0xcdd | RW | Default: 0x16c3 |
|---|---|---|
| [2:0] | 0x3 | Stage 3 calibration convergence speed in foreground mode (see Table 18). |
| [11:3] | 0xd8 | Reserved. |
| [13:12] | 0x1 | Stage 3 calibration randomization sequence in foreground mode (see Table 18). |

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| Address: 0xce1 | | RW | Default: 0xbbe8 |
|---|---|---|---|
| [7:0] | 0xe8 | Time allocated for calibration (see Table 18). | |
| [8] | 0x1 | Enable calibration (see Table 18). | |
| [9] | 0x1 | Enable foreground calibration (see Table 18). | |
| [10] | 0x0 | Reserved. | |
| [11] | 0x1 | Enable stg1 calibration (see Table 18). | |
| [12] | 0x1 | Enable stage 2 calibration (see Table 18). | |
| [13] | 0x1 | Enable stage 3 calibration (see Table 18). | |
| [14] | 0x0 | Background calibration parallel mode (see Table 18). | |
| [15] | 0x1 | Disable background calibration mode for all stages (see Table 18). | |

| Address: 0xce3 | | RW | Default: 0x0008 |
|---|---|---|---|
| [2:0] | 0x0 | Reserved. | |
| [3] | 0x1 | Clipping control to reserve some headroom for background calibration signals (active low). | |

| Address: 0xceb | | RW | Default: 0x0000 |
|---|---|---|---|
| [0] | 0x0 | Calibration status.<br>0: Not done<br>1: Done | |
| [7:1] | 0x0 | Reserved. | |

| Address: 0xcef | RO | Default: N/A |
|---|---|---|
| [0] | Calibration complete. | |
| [15:1] | Reserved. | |

## 0x800 - ADC_DSP

| Address: 0x805 | | RW | Default: 0x0000 |
|---|---|---|---|
| [1:0] | 0x0 | DSP Output clock divider:<br>00: Disabled (default)<br>01: Divide by 2<br>10: Divide by 4<br>11: N/A | |

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| Address: 0x807 | | RW | Default: 0x0300 |
|---|---|---|---|
| [0] | 0x0 | Interleaving Control. | |
| [6:1] | 0x0 | Reserved. | |
| [7] | 0x0 | Bypass gain correction.<br>0: Include gain correction<br>1: Bypass gain correction | |
| [8] | 0x1 | Bypass phase correction.<br>0: Include phase correction<br>1: Bypass phase correction | |
| [9] | 0x1 | Bypass digital down converter (DDC).<br>0: Include DDC<br>1: Bypass DDC | |
| [10] | 0x0 | Swap I and Q channels.<br>0: No swap<br>1: Swap | |
| [12:11] | 0x0 | A/Q data delay. | |
| [14:13] | 0x0 | B/I data delay. | |

| Address: 0x811 | | RW | Default: 0x0000 |
|---|---|---|---|
| [9:0] | 0x0 | IQ phase correction. Signed 10-bit two's complement number. Valid range from -511 to +511. | |
| [10] | 0x0 | Enable IQ phase correction.<br>0: Disable<br>1: Enable | |

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| Address: 0x813 | | RW | Default: 0x12a8 |
|---|---|---|---|
| [0] | 0x0 | Enable DDC Q path.<br>0: Disable<br>1: Enable | |
| [1] | 0x0 | Enable DDC I path.<br>0: Disable<br>1: Enable | |
| [3:2] | 0x2 | LO1 select:<br>0: Disabled<br>1: DDS1<br>2: DDS2<br>3: N/A | |
| [5:4] | 0x2 | LO2 select:<br>0: Disabled<br>1: DDS1<br>2: DDS2<br>3: N/A | |
| [6] | 0x0 | Low power mode.<br>0: Disable<br>1: Enable | |
| [7] | 0x1 | LO in low power mode:<br>0:fs/4<br>1: fs/2 | |
| [8] | 0x2 | IQ sum spectral inversion:<br>0: I*sin<br>1: I*sin inverted | |
| [9] | 0x0 | IQ sum spectral inversion:<br>0: Q*cos<br>1: Q*cos inverted | |
| [10] | 0x0 | IQ sum spectral inversion:<br>0: I*cos<br>1: I*cos inverted | |
| [11] | 0x0 | IQ sum spectral inversion:<br>0: Q*sin<br>1: Q*sin inverted | |
| [12] | 0x1 | IQ sum gain:<br>0: 1.0<br>1: 0.75 | |
| [14:13] | 0x0 | Output data select:<br>0: bypass<br>1: I and Q<br>2: IQ sum<br>3: LO | |

| Address: 0x815 | | RW | Default: 0x0000 |
|---|---|---|---|
| [15:0] | 0x0000 | DDS1 frequency: bits 15:0. | |

| Address: 0x817 | | RW | Default: 0x0000 |
|---|---|---|---|
| [15:0] | 0x0000 | DDS1 frequency: bits 31:16. | |

| Address: 0x819 | | RW | Default: 0x0000 |
|---|---|---|---|
| [15:0] | 0x0000 | DDS1 frequency: bits 47:32. | |

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| Address: 0x81b | | RW | Default: 0x0000 |
|---|---|---|---|
| [15:0] | 0x0000 | DDS2 frequency: bits 15:0. | |

| Address: 0x81d | | RW | Default: 0x0000 |
|---|---|---|---|
| [15:0] | 0x0000 | DDS2 frequency: bits 31:16. | |

| Address: 0x81f | | RW | Default: 0x0000 |
|---|---|---|---|
| [15:0] | 0x0000 | DDS2 frequency: bits 47:32. | |

| Address: 0x821 | | RW | Default: 0x0000 |
|---|---|---|---|
| [15:0] | 0x0000 | DDS1 phase offset. Signed 16-bit two's complement number. | |

| Address: 0x823 | | RW | Default: 0x0000 |
|---|---|---|---|
| [15:0] | 0x0000 | DDS2 phase offset. Signed 16-bit two's complement number. | |

| Address: 0x825 | | RW | Default: 0x0010 |
|---|---|---|---|
| [5:0] | 0x10 | DDS1 sine amplitude. Signed 6-bit two's complement number. | |

| Address: 0x827 | | RW | Default: 0x0000 |
|---|---|---|---|
| [5:0] | 0x0 | DDS1 sine offset. Signed 6-bit two's complement number. | |
| [11:6] | 0x0 | DDS1 cosine offset. Signed 6-bit two's complement number. | |

| Address: 0x829 | | RW | Default: 0x0000 |
|---|---|---|---|
| [5:0] | 0x0 | DDS2 sine offset. Signed 6-bit two's complement number. | |
| [11:6] | 0x0 | DDS2 cosine offset. Signed 6-bit two's complement number. | |
| [12] | 0x0 | DDS phase dither enable.<br>0: Disable<br>1: Enable | |
| [13] | 0x0 | DDS amplitude dither enable.<br>0: Disable<br>1: Enable | |

| Address: 0x82b | | RW | Default: 0x0000 |
|---|---|---|---|
| [0] | 0x0 | Load DDS1 frequency. Toggle to load DDS1 frequency. | |
| [1] | 0x0 | Load DDS2 frequency. Toggle to load DDS2 frequency. | |
| [2] | 0x0 | Load DDS1 phase. Toggle to load DDS1 phase offset. | |
| [3] | 0x0 | Load DDS2 phase. Toggle to load DDS2 phase offset. | |
| [5:4] | 0x0 | DDS mode:<br>0: sine<br>1: saw tooth<br>2: triangle<br>3: phase | |

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| Address: 0x82d | | RW | Default: 0x0130 |
|---|---|---|---|
| [3:0] | 0x0 | Reserved. | |
| [5:4] | 0x3 | Decimator pulse rate:<br>0: 1<br>1: 1/2<br>2: 1/4<br>3: N/A | |
| [8:6] | 0x4 | Reserved. | |

| Address: 0x82f | | RW | Default: 0x0147 |
|---|---|---|---|
| [0] | 0x1 | Bypass decimator.<br>0: Do not bypass<br>1: Bypass | |
| [1] | 0x1 | Enable decimator input path A.<br>0: Disable<br>1: Enable | |
| [2] | 0x1 | Enable decimator input path B.<br>0: Disable<br>1: Enable | |
| [3] | 0x0 | Enable decimator path C.<br>0: Disable<br>1: Enable | |
| [5:4] | 0x0 | Decimation factor:<br>0: 2x<br>1: 2x, interleaved ADC<br>2: 4x<br>3: 4x, interleaved ADC | |
| [7:6] | 0x1 | Filter 1 gain:<br>0: 1.0<br>1: 0.75<br>2: 0.5<br>3: 0.25 | |
| [9:8] | 0x1 | Filter 2 gain:<br>0: 1.0<br>1: 0.75<br>2: 0.5<br>3: 0.25 | |

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# REVISION HISTORY

| Version | Date | Comment |
|---|---|---|
| 1.0 | April 9, 2026 | Initial Release. |
| 2.0 | October 5, 2026 | Updated package drawing. |
| | | |

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